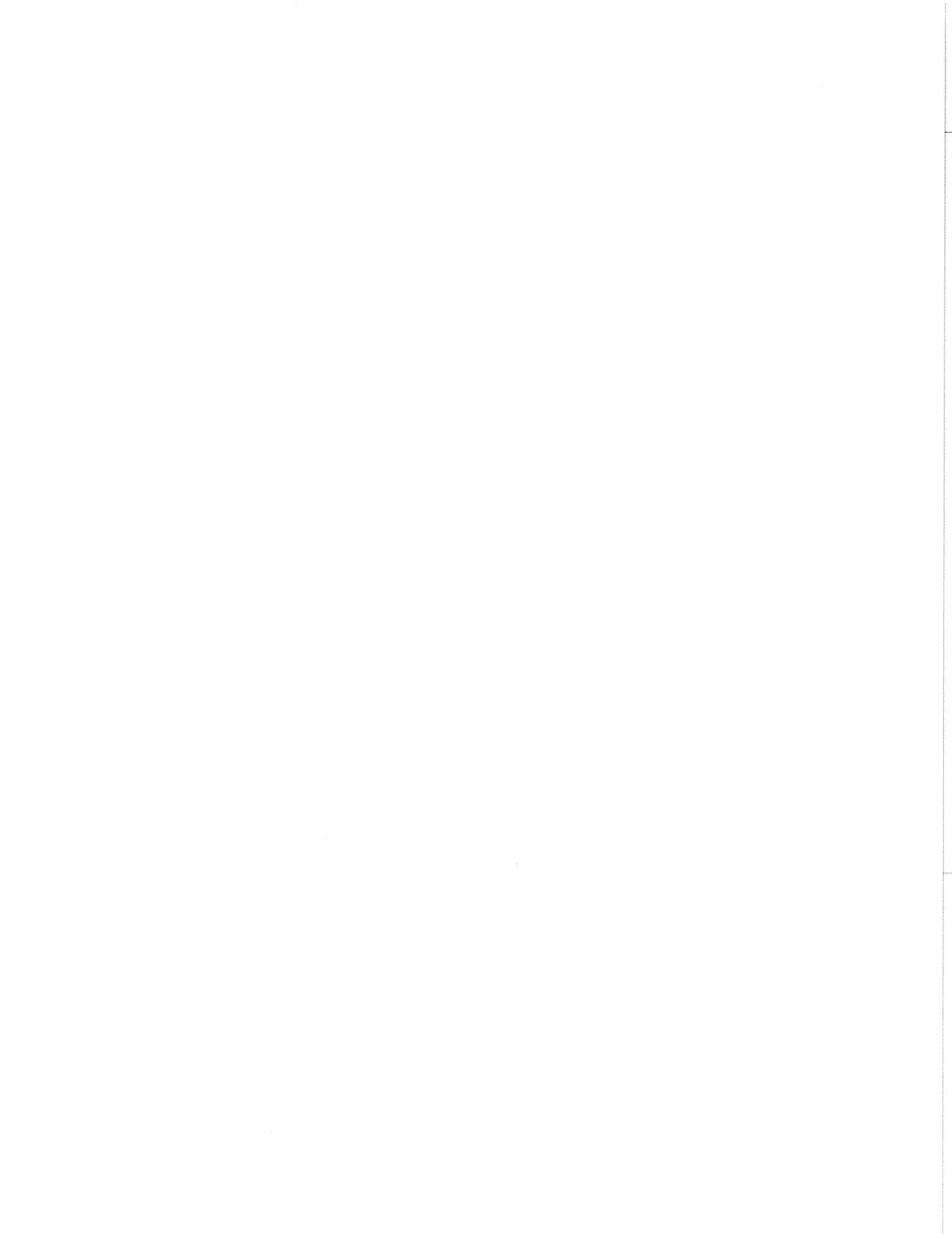


# Chapter Ten:

# Magnetically Coupled

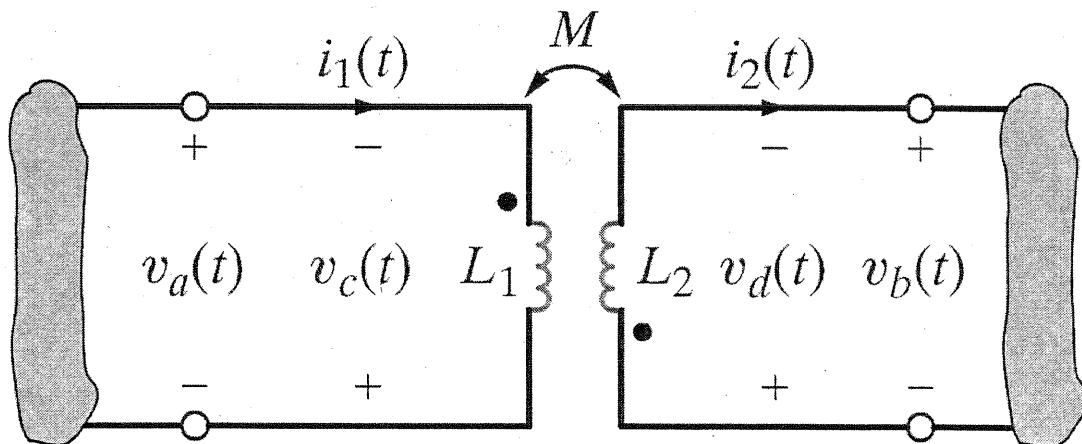
# Networks



**10.1** Given the network in Fig. P10.1,

(a) find the equations for  $v_a(t)$  and  $v_b(t)$ .

(b) find the equations for  $v_c(t)$  and  $v_d(t)$ .



**Figure P10.1**

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SOLUTION:

$$a) \quad v_a(t) = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt} \quad v_b(t) = -M \frac{di_1}{dt} - L_2 \frac{di_2}{dt}$$

$$b) \quad v_c(t) = -v_a(t) = -L_1 \frac{di_1}{dt} - M \frac{di_2}{dt}$$

$$v_d(t) = -v_b(t) = M \frac{di_1}{dt} + L_2 \frac{di_2}{dt}$$

10.2 Given the network in Fig. P10.2,

- (a) write the equations for  $v_a(t)$  and  $v_b(t)$ .
- (b) write the equations for  $v_c(t)$  and  $v_d(t)$ .

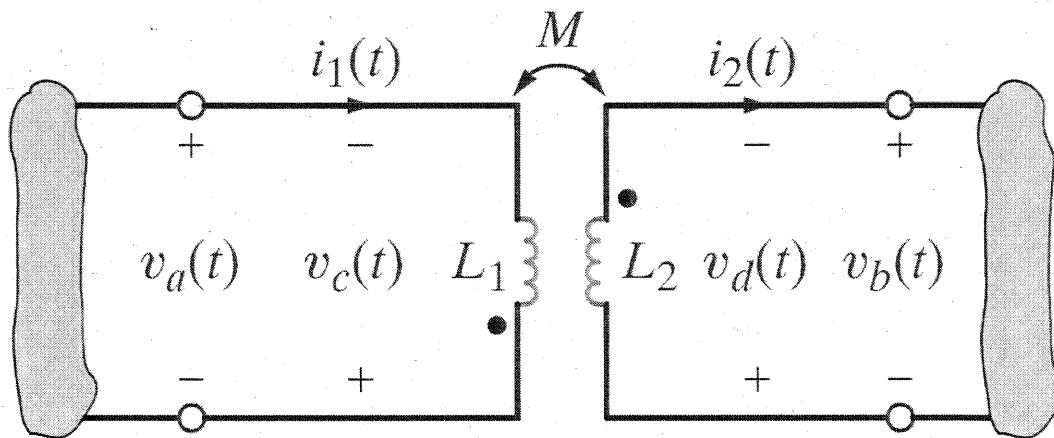


Figure P10.2

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SOLUTION:

$$a) \quad v_a(t) = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt} \quad v_b(t) = -M \frac{di_1}{dt} - L_2 \frac{di_2}{dt}$$

$$b) \quad v_c(t) = -v_a(t) = -L_1 \frac{di_1}{dt} - M \frac{di_2}{dt}$$

$$v_d(t) = -v_b(t) = M \frac{di_1}{dt} + L_2 \frac{di_2}{dt}$$

**10.3** Given the network in Fig. P10.3, **cs**

- (a) find the equations for  $v_a(t)$  and  $v_b(t)$ .
- (b) find the equations for  $v_c(t)$  and  $v_d(t)$ .

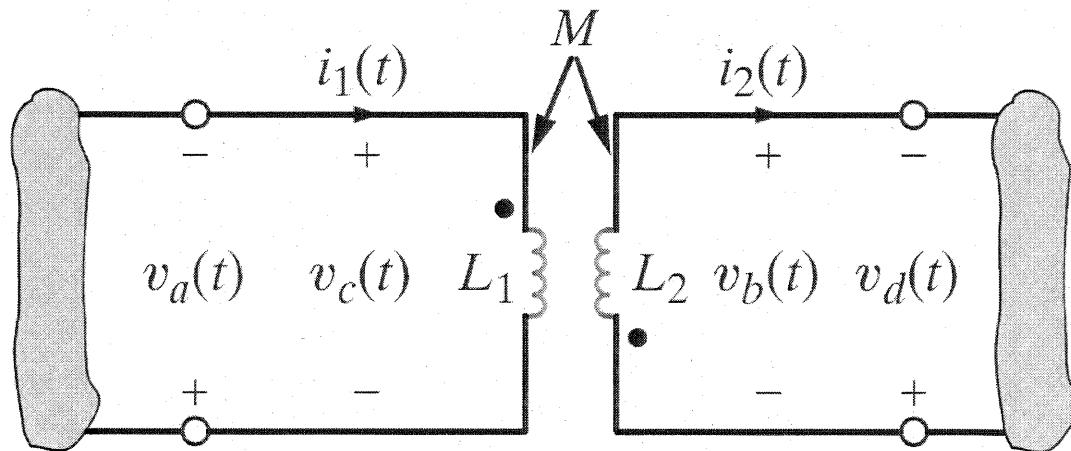


Figure P10.3

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SOLUTION:

$$2) \quad v_a(t) = -L_1 \frac{di_1}{dt} - M \frac{di_2}{dt} \quad v_b(t) = -M \frac{di_1}{dt} - L_2 \frac{di_2}{dt}$$

$$b) \quad v_c(t) = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt} \quad v_d(t) = M \frac{di_1}{dt} + L_2 \frac{di_2}{dt}$$

10.4 Given the network in Fig. P10.4,

- (a) write the equations for  $v_a(t)$  and  $v_b(t)$ .
- (b) write the equations for  $v_c(t)$  and  $v_d(t)$ .

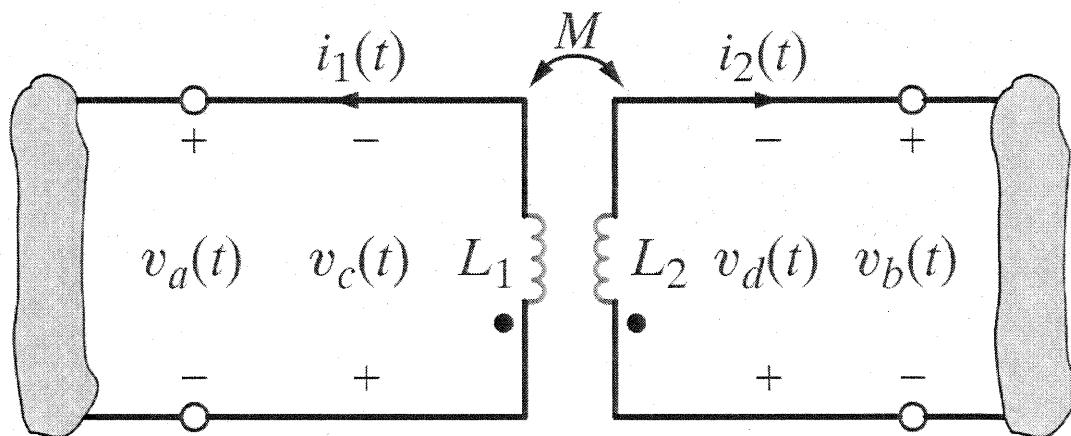


Figure P10.4

SOLUTION:

$$a) \quad v_a(t) = -L_1 \frac{di_1}{dt} - M \frac{di_2}{dt} \quad v_b(t) = -\frac{M di_1}{dt} - L_2 \frac{di_2}{dt}$$

$$b) \quad v_c(t) = L_1 \frac{di_1}{dt} + M \frac{di_2}{dt} \quad v_d(t) = M \frac{di_1}{dt} + L_2 \frac{di_2}{dt}$$

- 10.5 Find the voltage gain  $V_o/V_s$  of the network shown in Fig. P10.5. CS

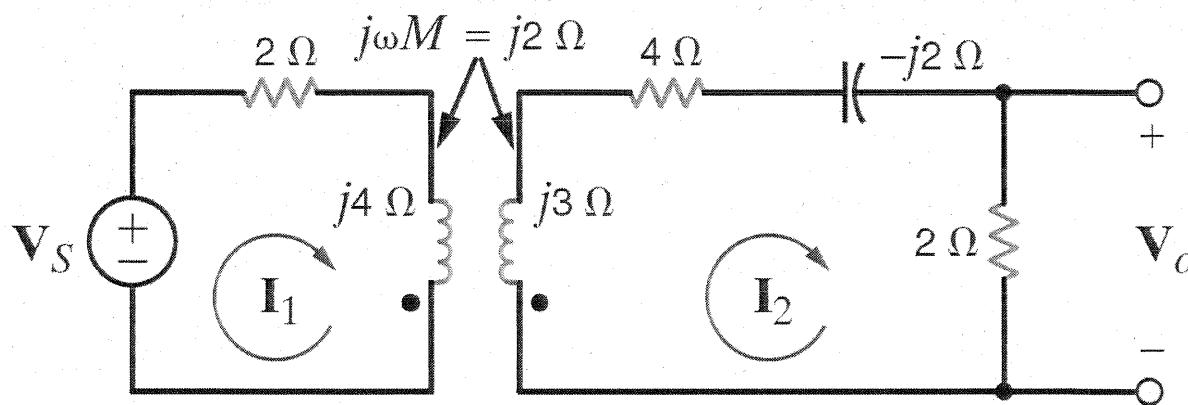


Figure P10.5

SOLUTION:

$$V_s = I_1(z + j4) - j2I_2 \quad & \quad 0 = -j2I_1 + I_2(6 + j1)$$

$$V_o = 2I_2 \quad \rightarrow \quad I_1 = \frac{V_s + j2I_2}{z + j4} \quad 0 = \frac{-j2(V_s + j2I_2)}{z + j4} + I_2(6 + j1)$$

$$I_2 = \frac{jV_s}{6 + j13} \quad V_o = \frac{j2V_s}{6 + j13} \quad \frac{V_o}{V_s} = \frac{j2}{6 + j13}$$

$$\boxed{\frac{V_o}{V_s} = 0.140 \angle 24.8^\circ}$$

10.6 Find  $V_o$  in the network in Fig. P10.6.

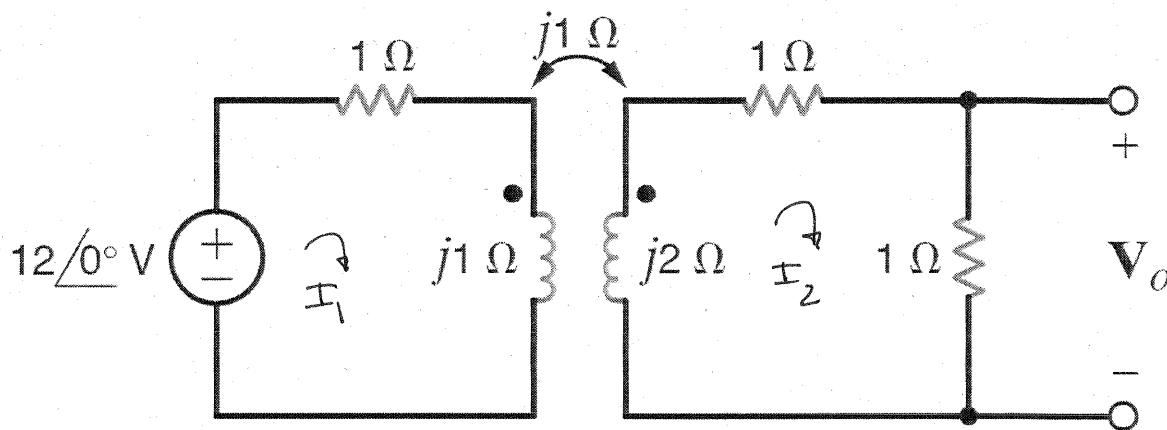


Figure P10.6

SOLUTION:

$$12 \angle 0^\circ = I_1(1+j1) - j1 I_2 \quad \& \quad 0 = -j1 I_1 + I_2(z+j2) \quad V_o = I_2(z)$$

$$\begin{bmatrix} 1+j1 & -j1 \\ -j1 & z+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 12 \\ 0 \end{bmatrix} \Rightarrow I_2 = 2.91 \angle 14.0^\circ \text{ A}$$

$$V_o = 2.91 \angle 14.0^\circ \text{ V}$$

10.7 Given the network in Fig. P10.7, find  $V_o$ . **PSV**

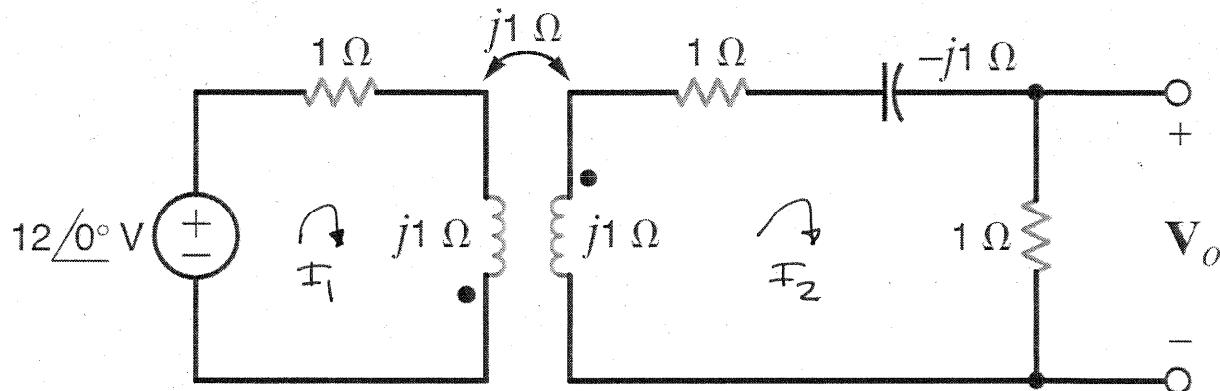


Figure P10.7

SOLUTION:

$$12\angle 0^\circ = I_1(1+j1) + j1 I_2 \quad \& \quad 0 = j1 I_1 + E_2(z) \quad \& \quad V_o = 1I_2$$

$$\begin{bmatrix} 1+j1 & j1 \\ j1 & 2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 12 \\ 0 \end{bmatrix} \Rightarrow I_2 = 3.32 \angle -124^\circ \text{ A}$$

$V_o = 3.32 \angle -124^\circ \text{ V}$

10.8 Find  $V_o$  in the circuit in Fig. P10.8. [cs]

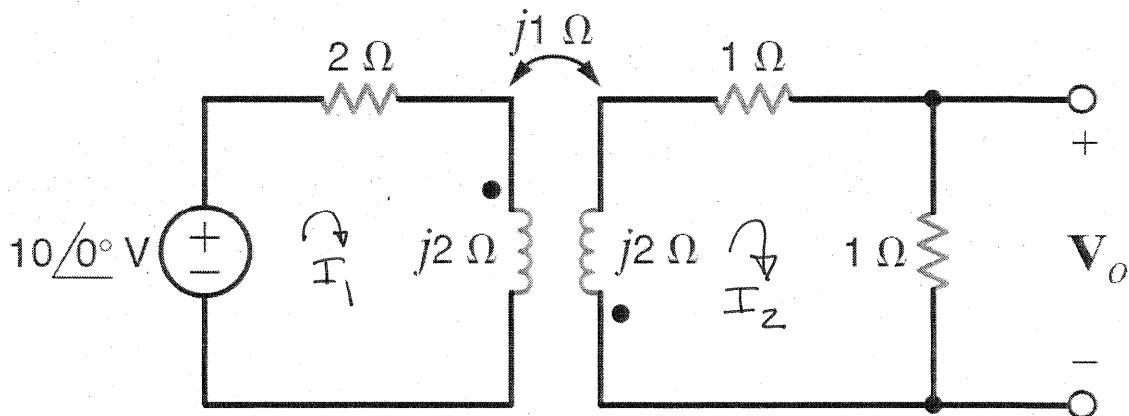


Figure P10.8

SOLUTION:

$$10\angle 0^\circ = I_1(2+j2) + j1I_2 \quad \text{and} \quad j1I_1 + I_2(2+j2) = 0 \quad \text{and} \quad V_o = (1)I_2$$

$$\begin{bmatrix} 2+j2 & j1 \\ j1 & 2+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 10 \\ 0 \end{bmatrix} \Rightarrow I_2 = 1.24 \angle -173^\circ \text{ A}$$

$V_o = 1.24 \angle -173^\circ \text{ V}$

- 10.9 Find the voltage gain  $V_o/V_s$  of the network shown in Fig. P10.9.

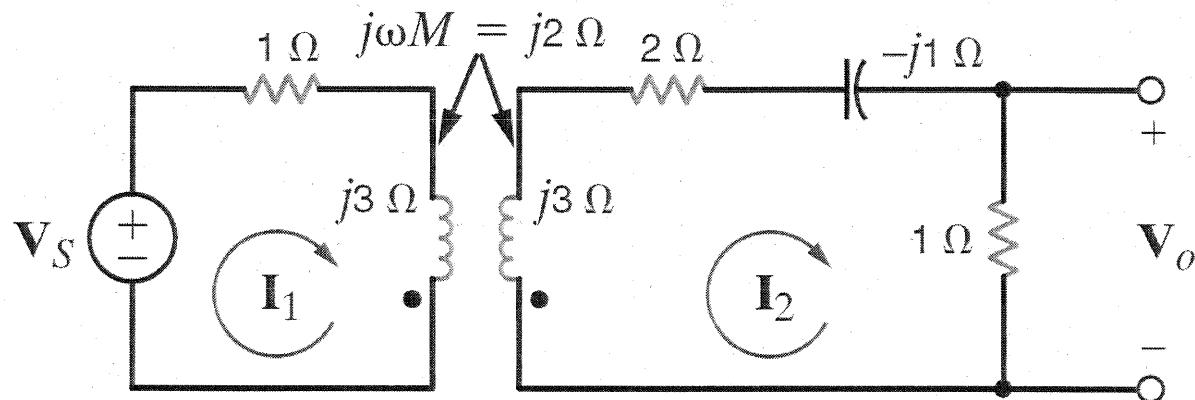


Figure P10.9

SOLUTION:

$$V_o = I_2 \quad \& \quad V_s = I_1(1+j3) - j2I_2 \quad \& \quad -j2I_1 + I_2(3+j2) = 0$$

$$\text{Let } V_s = 1 \angle 0^\circ \text{ V. Then } \frac{V_o}{V_s} = V_o$$

$$\begin{bmatrix} 1+j3 & -j2 \\ -j2 & 3+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \Rightarrow I_2 = 0.181 \angle 5.19^\circ \text{ A}$$

$\frac{V_o}{V_s} = 0.181 \angle 5.19^\circ$
---

10.10 Find  $V_o$  in the network in Fig. P10.10.

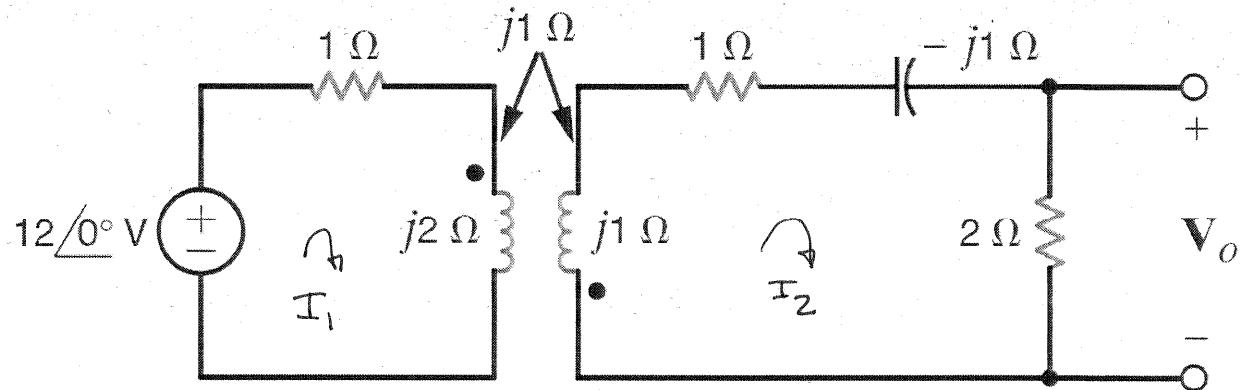


Figure P10.10

SOLUTION:

$$12\angle 0^\circ = I_1(1+j2) + j1I_2 \quad \& \quad j1I_1 + I_2(3) = 0 \quad \& \quad V_o = 2I_2$$

$$\begin{bmatrix} 1+j2 & j1 \\ j1 & 3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 12 \\ 0 \end{bmatrix} \Rightarrow I_2 = 1.66 \angle -146^\circ \text{ A}$$

$$V_o = 3.33 \angle -146^\circ \text{ V}$$

10.11 Find  $V_o$  in the network in Fig. P10.11.

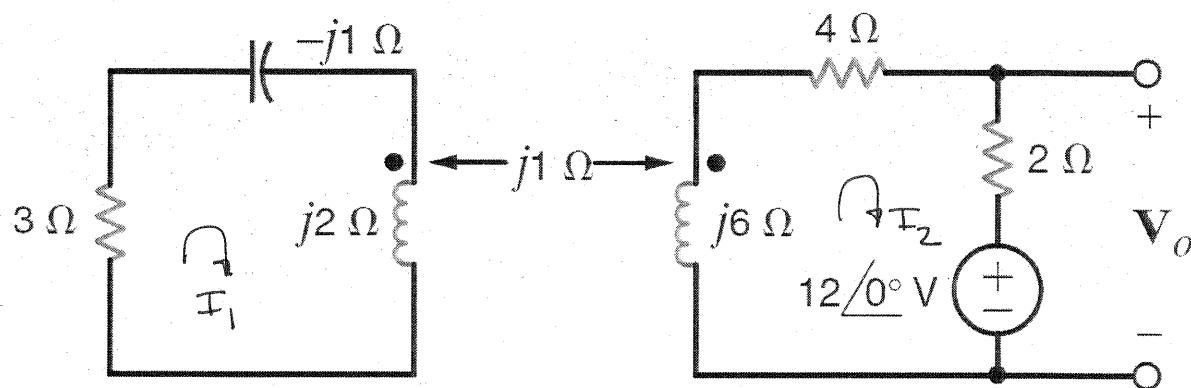


Figure P10.11

SOLUTION:

$$I_1(3+j1) - jI_2 = 0 \quad \& \quad -12\angle 0^\circ = -jI_1 + I_2(6+j6) \quad \& \quad V_o = 12\angle 0^\circ + 2I_2$$

$$\begin{bmatrix} 3+j1 & -j1 \\ -j1 & 6+j6 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 \\ -12 \end{bmatrix} \Rightarrow I_2 = 1.39 \angle 137^\circ \text{ A}$$

$$V_o = 10.15 \angle 10.8^\circ \text{ V}$$

10.12 Find  $V_o$  in the circuit in Fig. P10.12.

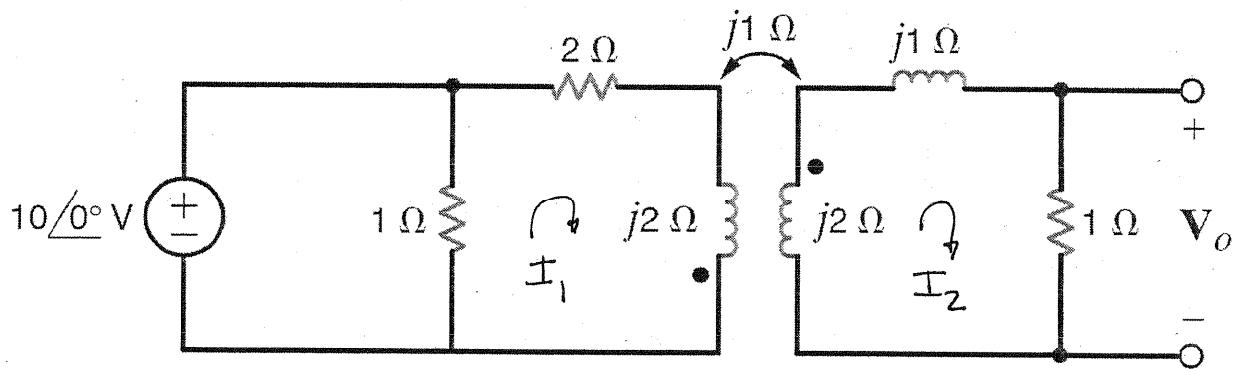


Figure P10.12

SOLUTION:

$$10 \angle 0^\circ = I_1(z + j2) + j1 I_2 \quad 0 = j1 I_1 + I_2(1 + j3) \quad V_o = 1 I_2$$

$$\begin{bmatrix} z + j2 & j1 \\ j1 & 1 + j3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 10 \\ 0 \end{bmatrix} \Rightarrow I_2 = 1.17 \angle 159^\circ \text{ A}$$

$V_o = 1.17 \angle 159^\circ \text{ V}$

10.13 Find  $V_o$  in the network in Fig. P10.13.

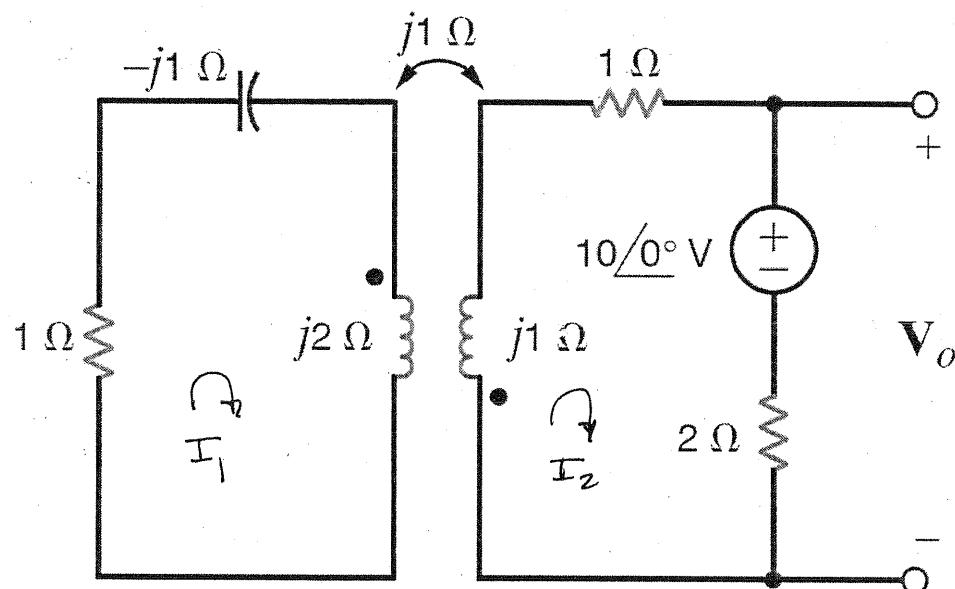


Figure P10.13

SOLUTION:

$$I_1(1+j1) + jI_2 = 0 \quad \& \quad -10\angle 0^\circ = jI_1 + I_2(3+j1) \quad \& \quad V_o = 2I_2 + 10\angle 0^\circ$$

$$\begin{bmatrix} 1+j1 & j1 \\ j1 & 3+j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 0 \\ -10 \end{bmatrix} \Rightarrow I_2 = 2.83 \angle 172^\circ \text{ A}$$

$V_o = 4.47 \angle 10.3^\circ \text{ V}$

10.14 Find  $V_o$  in the network in Fig. P10.14.

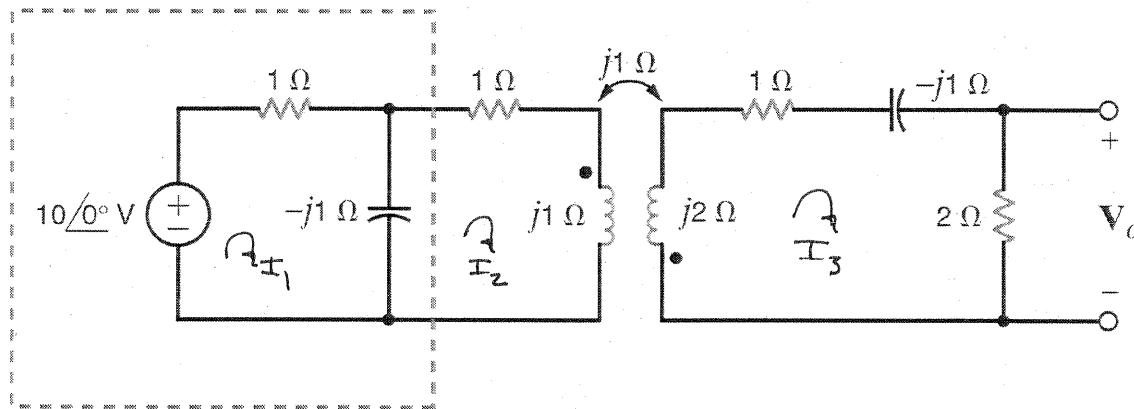


Figure P10.14

SOLUTION:

$$I_o = I_1(1-j1) + j(I_2 + I_3) \neq jI_1 + I_2(1) + jI_3 = 0 \neq jI_2 + I_3(3+j1) = 0$$

$$\begin{bmatrix} 1-j1 & j1 & 0 \\ j1 & 1 & j1 \\ 0 & j1 & 3+j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} I_o \\ 0 \\ 0 \end{bmatrix} \Rightarrow \begin{aligned} V_o &= 2I_3 \\ I_3 &= 1.21 \angle -166^\circ \text{ A} \\ V_o &= 2.42 \angle -166^\circ \text{ V} \end{aligned}$$

10.15 Find  $\mathbf{V}_o$  in the network in Fig. P10.15.

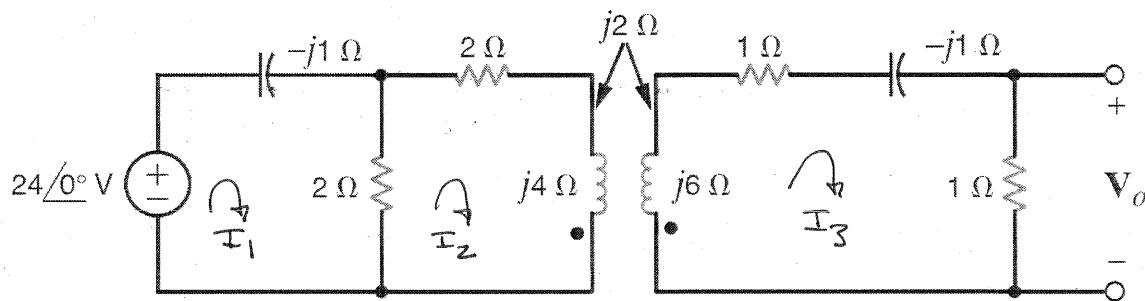


Figure P10.15

SOLUTION:

$$24\angle 0^\circ = I_1(z-j1) - 2I_2 \neq -2I_1 + I_2(4+j4) - j2I_3 = 0 \neq -j2I_2 + I_3(z+j5) = 0$$

$$\begin{bmatrix} 2-j1 & -2 & 0 \\ -2 & 4+j4 & -j2 \\ 0 & -j2 & 2+j5 \end{bmatrix} \begin{bmatrix} z+j \\ 0 \\ 0 \end{bmatrix} \Rightarrow \begin{aligned} V_o &= (1)I_3 \\ I_3 &= 2.17 \angle 5.19^\circ A \\ V_o &= 2.17 \angle 5.19^\circ V \end{aligned}$$

10.16 Find  $V_o$  in the circuit in Fig. P10.16.

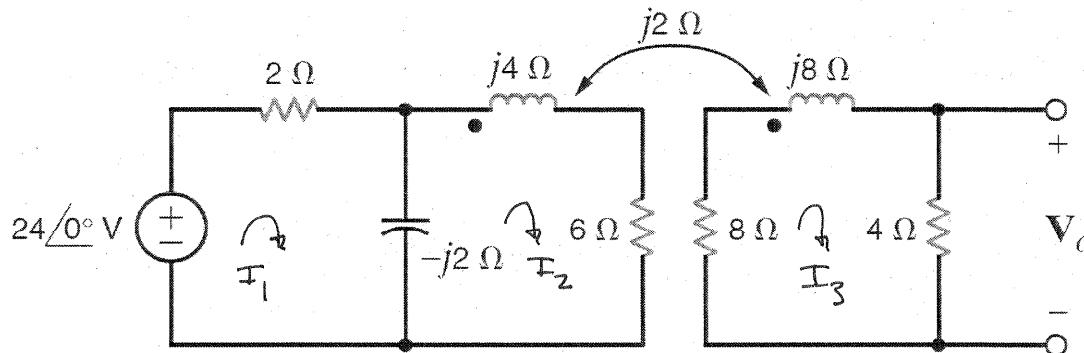


Figure P10.16

SOLUTION:

$$24\angle 0^\circ = I_1(z - j2) + j2I_2 \quad \& \quad j2I_1 + I_2(6 + j2) + j2I_3 = 0 \quad \& \quad j2I_2 + I_3(12 + j8) = 0$$

$$\begin{bmatrix} z - j2 & j2 & 0 \\ j2 & 6 + j2 & j2 \\ 0 & j2 & 12 + j8 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 24 \\ 0 \\ 0 \end{bmatrix} \Rightarrow \begin{aligned} I_3 &= 0.303 \angle 170^\circ \text{ A} \\ V_o &= 4I_3 \end{aligned}$$

$$V_o = 1.21 \angle 170^\circ \text{ V}$$

10.17 Find  $V_o$  in the network in Fig. P10.17.

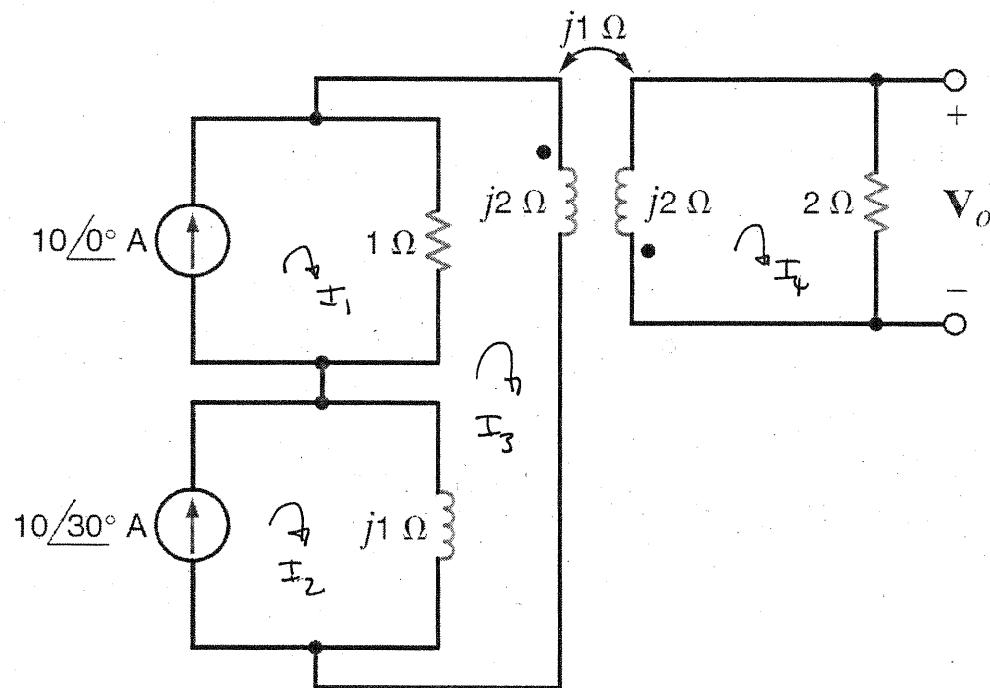


Figure P10.17

SOLUTION:

$$I_1 = 10 \angle 0^\circ \text{ A} \quad I_2 = 10 \angle 30^\circ \text{ A} \quad I_3 (1+j3) - I_1 - jI_2 + jI_4 = 0$$

$$\text{yields, } I_3 (1+j3) + jI_4 = I_1 + jI_2 = 5 + j8.66$$

$$\text{and } jI_3 + I_4 (2+j2) = 0 \quad \& \quad V_o = 2I_4$$

$$\begin{bmatrix} 1+j3 & j1 \\ j1 & (2+j2) \end{bmatrix} \begin{bmatrix} I_3 \\ I_4 \end{bmatrix} = \begin{bmatrix} 5+j8.66 \\ 0 \end{bmatrix} \Rightarrow I_4 = 1.17 \angle -141^\circ \text{ A}$$

$$V_o = 2.34 \angle -141^\circ \text{ V}$$

10.18 Find  $I_o$  in the circuit in Fig. P10.18.

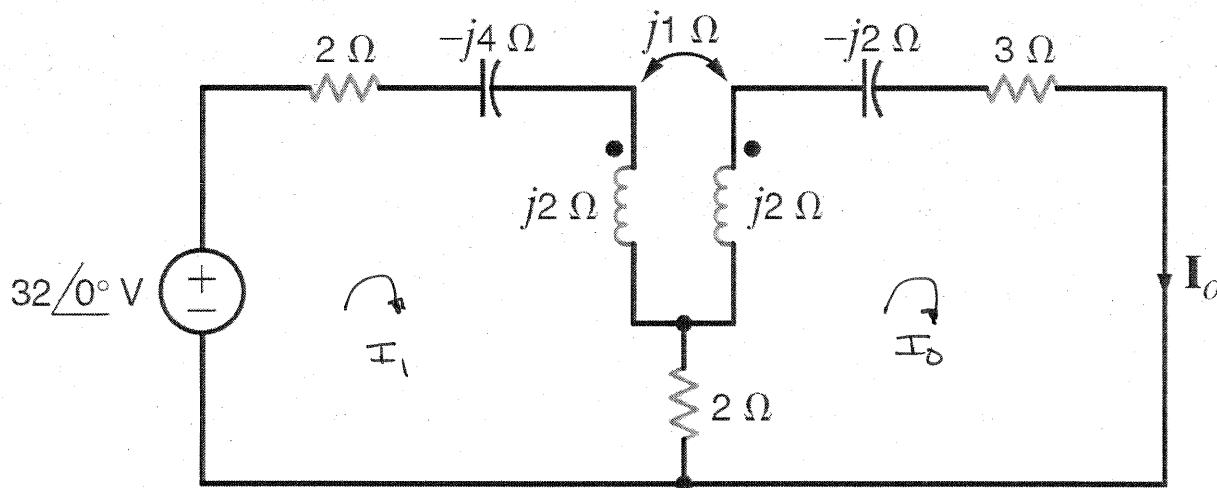


Figure P10.18

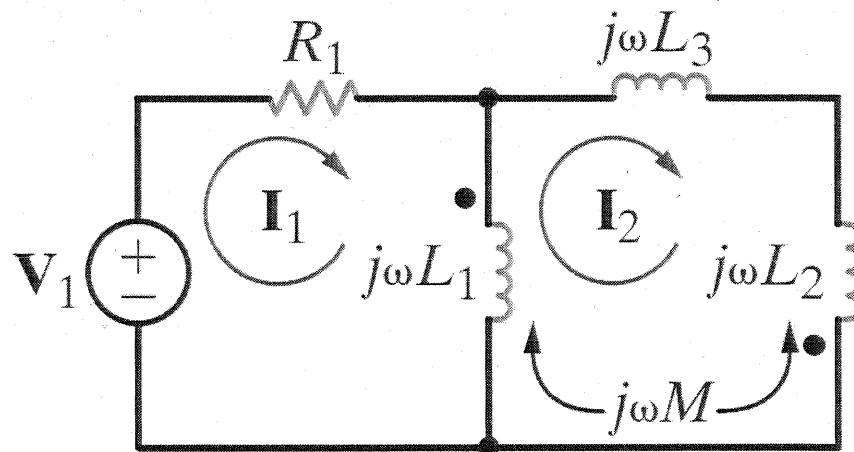
SOLUTION:

$$32\angle 0^\circ = I_1(2-j4+j2+z) - jI_o - 2I_o \Rightarrow 32\angle 0^\circ = I_1(4-j2) + I_o(-2-j1)$$

$$0 = -I_1(j1) - 2I_1 + I_o(z+j2-j2+3) \Rightarrow 0 = I_1(-z-j1) + I_o(5)$$

$$\begin{bmatrix} 4-j2 & -2-j1 \\ -2-j1 & 5 \end{bmatrix} \begin{bmatrix} I_1 \\ I_o \end{bmatrix} = \begin{bmatrix} 32 \\ 0 \end{bmatrix} \Rightarrow I_o = 3.25 \angle 66.0^\circ \text{ A}$$

- 10.19 Write the mesh equations for the network in Fig. P10.19.



**Figure P10.19**

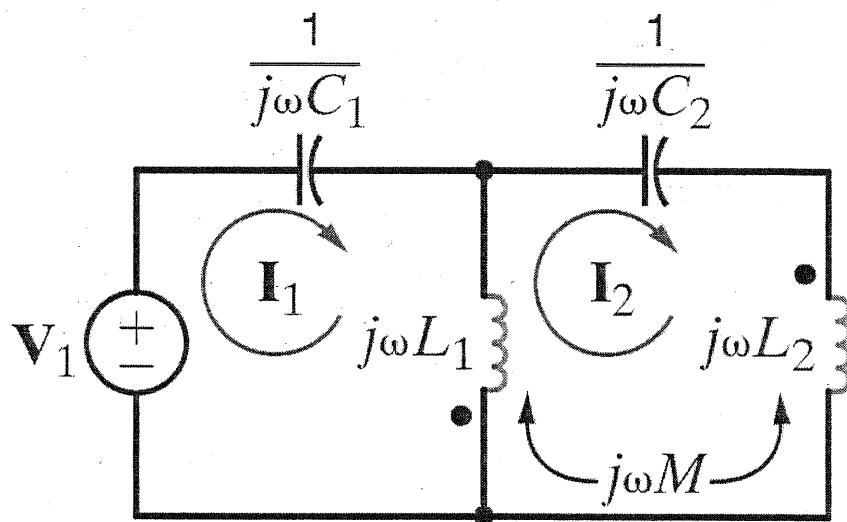
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SOLUTION:

$$V_1 = I_1(R_1 + j\omega L_1) - j\omega M I_2$$

$$0 = -j\omega M I_1 + I_2(j\omega)(L_1 + L_2 + L_3)$$

**10.20** Write the mesh equations for the network in Fig. P10.20.



**Figure P10.20**

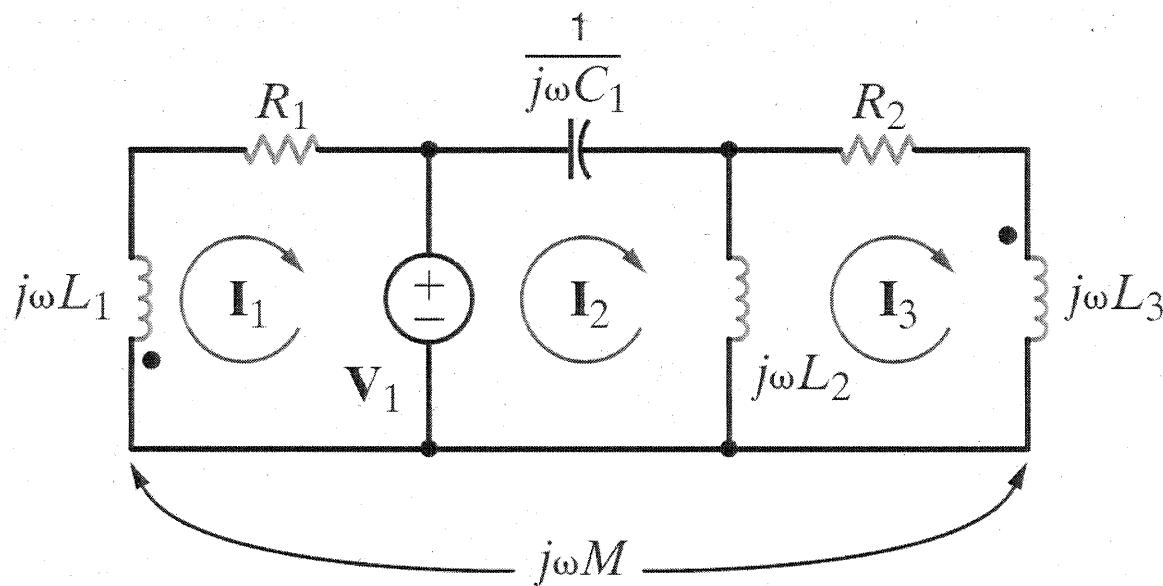
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SOLUTION:

$$V_1 = I_1 \left( j\omega L_1 - \frac{1}{j\omega C_1} \right) - j\omega M I_2 - j\omega L_1 I_2$$

$$0 = -j\omega M I_1 + I_2 \left( j\omega (L_1 + L_2) - \frac{1}{j\omega C_2} \right) - j\omega L_1 I_1$$

- 10.21** Write the mesh equations for the network shown in Fig. P10.21. **cs**



**Figure P10.21**

**SOLUTION:**

$$-V_1 = I_1 (R_1 + j\omega L_1) + j\omega M I_3$$

$$V_1 = I_2 (j\omega L_2 - j\omega C_1) - j\omega L_2 I_3$$

$$0 = -j\omega L_2 I_2 + I_3 (R_2 + j\omega (L_2 + L_3)) + j\omega M I_1$$

- 10.22 Write the mesh equations for the network shown in Fig. P10.22.

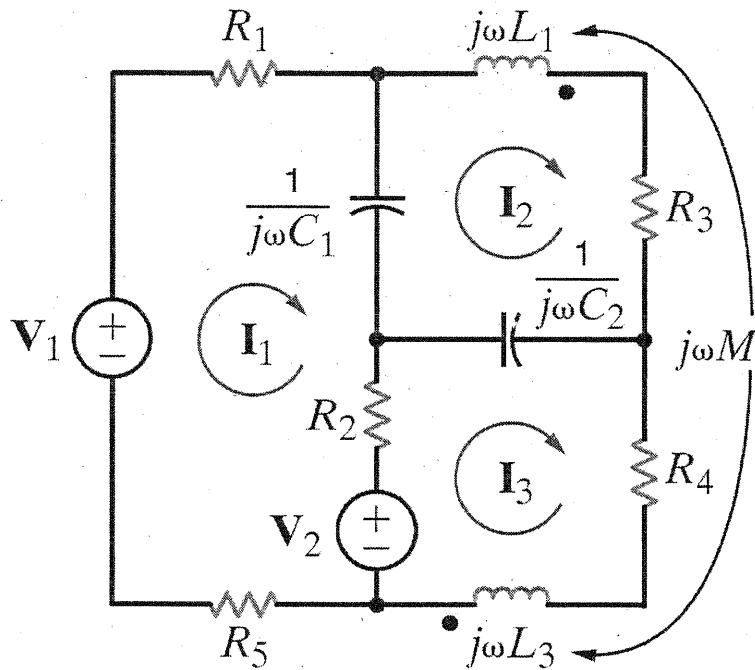


Figure P10.22

SOLUTION:

$$V_1 - V_2 = I_1(R_1 + R_2 + R_5 - j/\omega C_1) + I_2(+j/\omega C_1) - R_2 I_3$$

$$0 = \frac{j}{\omega C_1} I_1 + I_2 \left( R_3 + j \left[ \omega L_1 - \frac{1}{\omega C_1} - \frac{1}{\omega C_2} \right] \right) + j \omega M I_3 + \frac{j}{\omega C_2} I_3$$

$$V_2 = -R_2 I_1 + I_2 \left( j \omega M + \frac{j}{\omega C_2} \right) + I_3 \left( R_2 + R_4 + j \omega L_3 - \frac{j}{\omega C_2} \right)$$

- 10.23** Write the mesh equations for the network in Fig. P10.23.

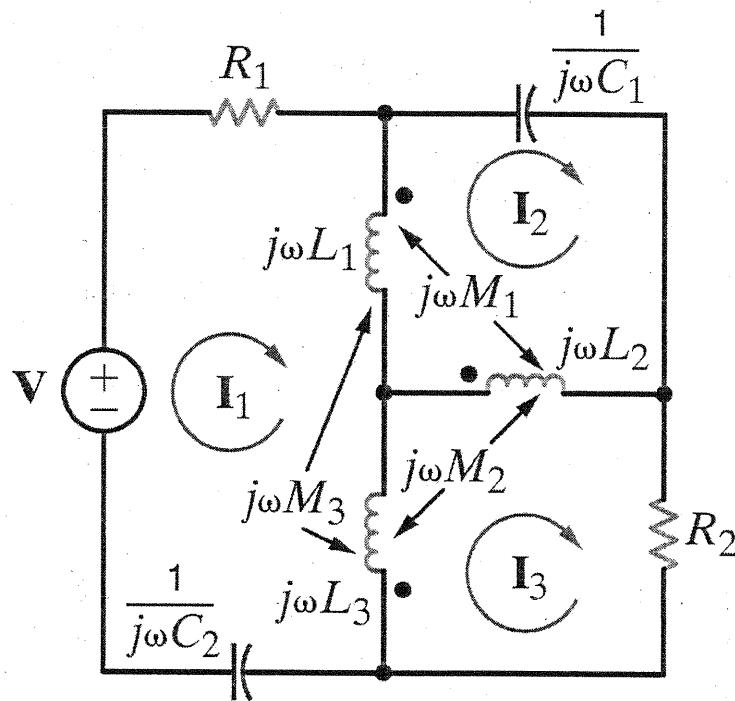


Figure P10.23

**SOLUTION:**

$$V = I_1 \left[ R_1 + j(\omega L_1 + \omega L_3 - \frac{1}{j\omega C_2}) \right] - j\omega L_1 I_2 - j\omega L_3 I_3 + j\omega M_3 (I_3 - I_1) + j\omega M_3 (I_2 - I_1) + j\omega M_1 (I_3 - I_2) + j\omega M_2 (I_2 - I_3) \quad (\text{eq. 1})$$

$$0 = I_2 \left[ j(\omega L_1 + \omega L_2 - \frac{1}{j\omega C_1}) \right] - j\omega L_1 I_1 - j\omega L_2 I_3 + j\omega M_1 (I_2 - I_1) + j\omega M_1 (I_2 - I_3) + j\omega M_2 (I_1 - I_3) + j\omega M_3 (I_1 - I_3) \quad (\text{eq. 2})$$

$$0 = I_3 \left[ R_2 + j\omega (L_2 + L_3) \right] - j\omega L_2 I_2 - j\omega L_3 I_1 + j\omega M_1 (I_1 - I_2) + j\omega M_2 (I_3 - I_1) + j\omega M_3 (I_1 - I_2) + j\omega M_2 (I_3 - I_2) \quad (\text{eq. 3})$$

$$V = I_1 \left[ F_1 + j \left( \omega L_1 + \omega L_3 - 2\omega M_3 - \frac{1}{\omega C_2} \right) + I_2 \left[ j\omega (M_3 + M_2 - M_1 - L_1) \right] \right. \\ \left. + I_3 \left[ M_1 - M_2 + M_3 - L_3 \right] j\omega \right]$$

$$0 = I_1 \left[ j\omega (M_2 + M_3 - M_1 - L_1) \right] + F_2 \left[ j \left( \omega (L_1 + L_2 + 2M_1) - \frac{1}{\omega C_1} \right) \right] \\ + I_3 \left[ j\omega (-M_1 - M_2 - M_3 - L_2) \right]$$

$$0 = I_1 \left[ j\omega (M_1 - M_2 + M_3 - L_3) \right] + I_2 \left[ j\omega (-M_1 - M_2 - M_3 - L_2) \right] \\ + I_3 \left[ R_2 + j\omega (L_2 + L_3 + 2M_2) \right]$$

10.24 Find  $V_o$  in the network in Fig. P10.24.

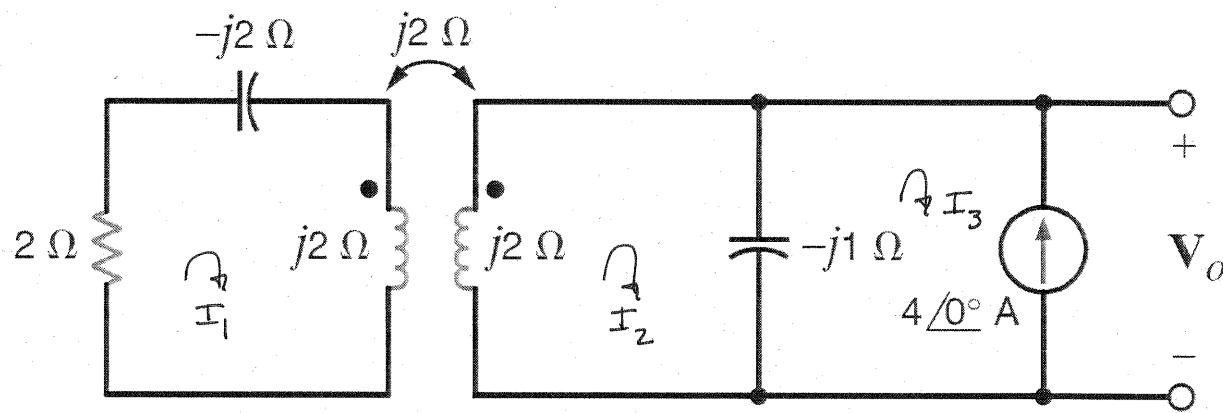


Figure P10.24

SOLUTION:

$$0 = I_1(z) - j2I_2 \quad \& \quad 0 = -j2I_1 + I_2(j1) + j1I_3 \quad \& \quad I_3 = -4\angle 0^\circ A$$

$$\begin{bmatrix} z & -j2 & 0 \\ -j2 & j1 & j1 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -4 \end{bmatrix}$$

$$V_o = -j1(I_2 - I_3)$$

$$I_2 = 0.8 + j1.6 A$$

$$I_3 = -4 + j0 A$$

$$V_o = 5.06 \angle -71.6^\circ V$$

10.25 Find  $V_o$  in the network in Fig. P10.25.

CS

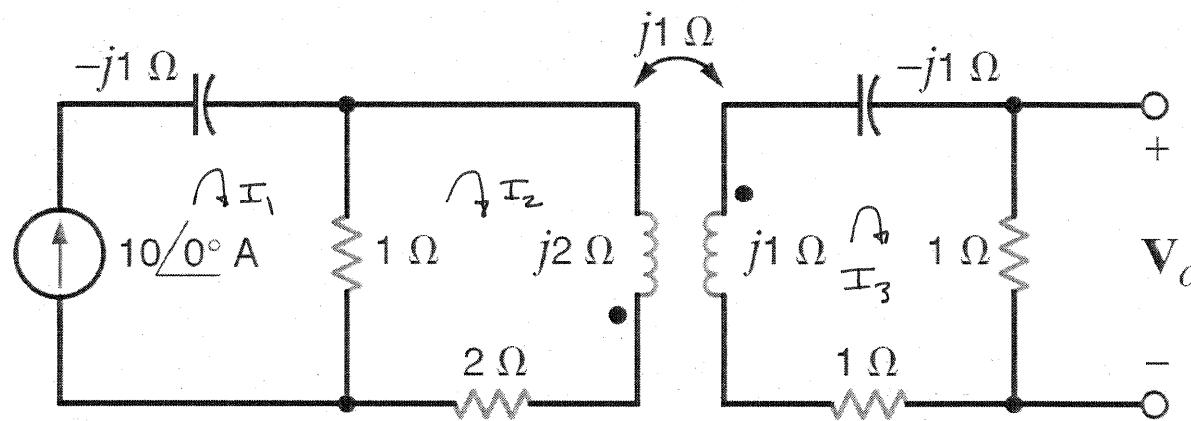


Figure P10.25

SOLUTION:

$$I_1 = 10 \angle 0^\circ \text{ A} \quad \& \quad I_2 (3+j2) - I_1 + jI_3 = 0 \quad \& \quad 0 = jI_2 + I_3 [z]$$

$$\begin{bmatrix} -1 & 3+j2 & j1 \\ 0 & j1 & z \\ 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$$

$$V_o = (1)I_3$$

$$I_3 = 1.24 \angle -120^\circ \text{ A}$$

$$V_o = 1.24 \angle -120^\circ \text{ V}$$

10.26 Find  $V_o$  in the network in Fig. P10.26.

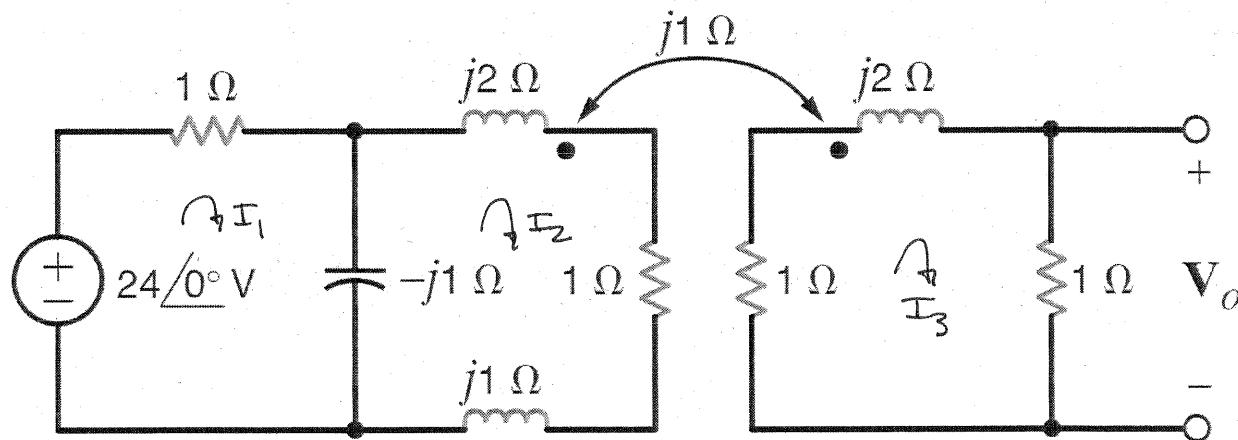


Figure P10.26

SOLUTION:

$$24 \angle 0^\circ = I_1(1-j1) + jI_2 \quad \text{and} \quad jI_1 + I_2(1+j2) - jI_3 = 0 \quad V_o = (1)I_3$$

$$0 = -jI_2 + I_3(z + j2)$$

$$\begin{bmatrix} 1-j1 & j1 & 0 \\ j1 & 1+j2 & -j1 \\ 0 & -j1 & 2+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 24 \\ 0 \\ 0 \end{bmatrix} \Rightarrow I_3 = 2.10 \angle -52.1^\circ \text{ A}$$

$$V_o = 2.10 \angle -52.1^\circ \text{ V}$$

10.27 Find  $V_o$  in the network in Fig. P10.27.

PSV

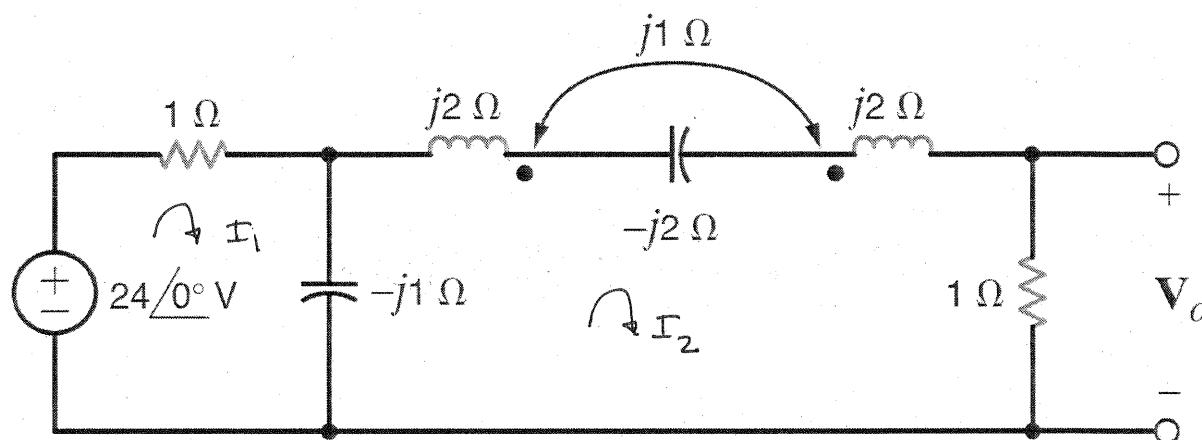


Figure P10.27

SOLUTION:

$$z + \angle 0^\circ = I_1(1-j1) + j1 I_2 \quad \text{and} \quad V_o = (1) I_2$$

$$0 = I_2(1+j1) + j1 I_1 - j1 I_2 - j1 I_2 \Rightarrow jI_1 + I_2(1-j1) = 0$$

$$\begin{bmatrix} 1-j1 & j1 \\ j1 & 1-j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} z + \\ 0 \end{bmatrix} \quad V_o = (1) I_2$$

$$I_2 = 10.7 \angle -26.6^\circ \text{ A}$$

$$V_o = 10.7 \angle -26.6^\circ \text{ V}$$

10.28 Find  $V_o$  in the network in Fig. P10.28.

CS

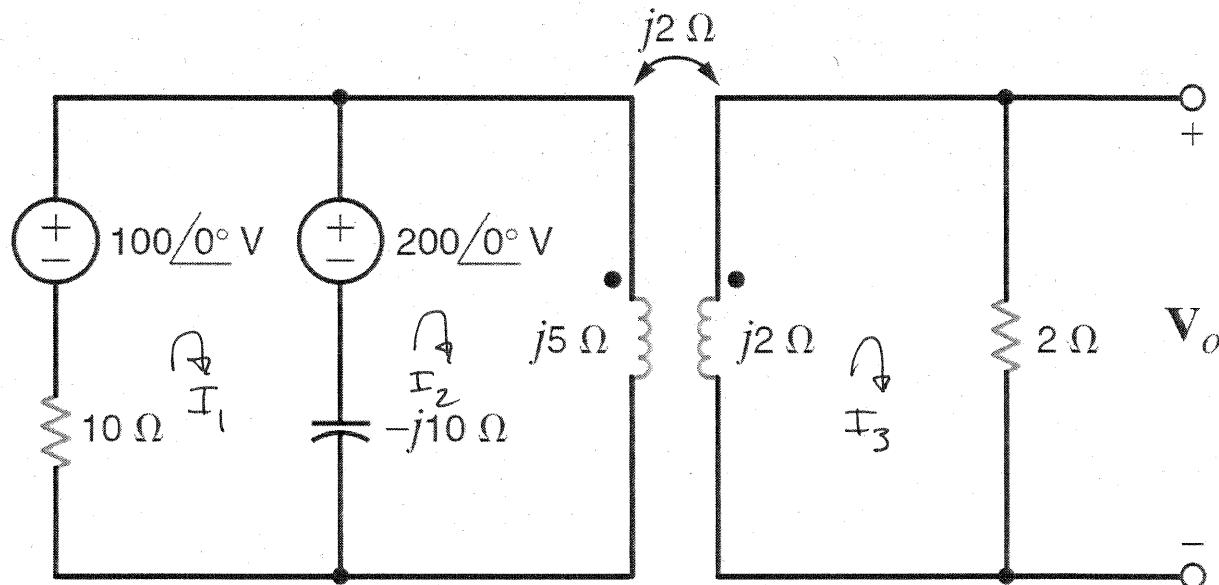


Figure P10.28

SOLUTION:

$$100 \angle 0^\circ - 200 \angle 0^\circ = I_1(10 - j10) + j10 I_2 = -100 \angle 0^\circ$$

$$200 \angle 0^\circ = j10 I_1 + I_2(-j5) - j2 I_3$$

$$0 = -j2 I_2 + I_3(2 + j2)$$

$$V_o = 2 I_3$$

$$\begin{bmatrix} 10 - j10 & j10 & 0 \\ j10 & -j5 & -j2 \\ 0 & -j2 & 2 + j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} -100 \\ 200 \\ 0 \end{bmatrix}$$

$$I_3 = 18.4 \angle 72.9^\circ A$$

$$V_o = 36.8 \angle 72.9^\circ V$$

10.29 Find  $\mathbf{V}_o$  in the network in Fig. P10.29.

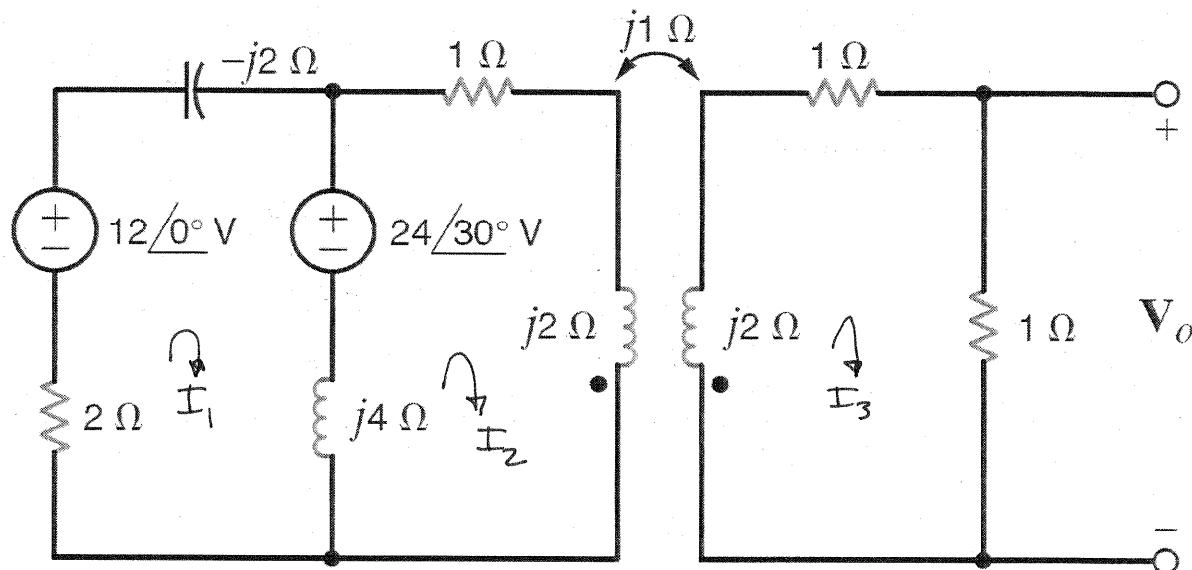


Figure P10.29

SOLUTION:

$$12 \angle 0^\circ - 24 \angle 30^\circ = I_1 (z + j2) - j4 I_2 = -8.78 - j12$$

$$24 \angle 30^\circ = -j4 I_1 + I_2 (1 + j6) - j I_3 \quad I_3 (1) = V_o$$

$$0 = -j I_2 + I_3 (z + j2)$$

$$\begin{bmatrix} z + j2 & -j4 & 0 \\ -j4 & 1 + j6 & -j1 \\ 0 & -j1 & z + j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} -8.78 - j12 \\ 20.8 + j12 \\ 0 \end{bmatrix}$$

$$I_3 = 1.63 \angle 6.49^\circ \text{ A}$$

$$V_o = 1.63 \angle 6.49^\circ \text{ V}$$

10.30 Find  $V_o$  in the network in Fig. P10.30.

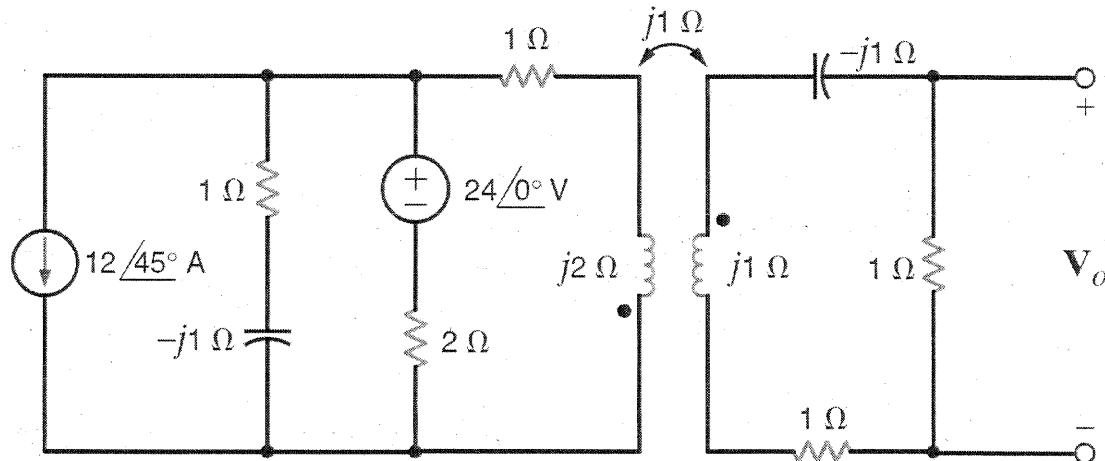
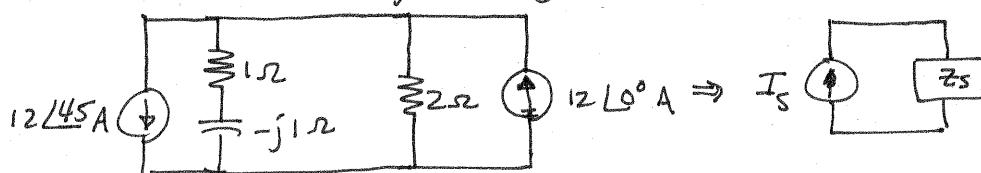


Figure P10.30

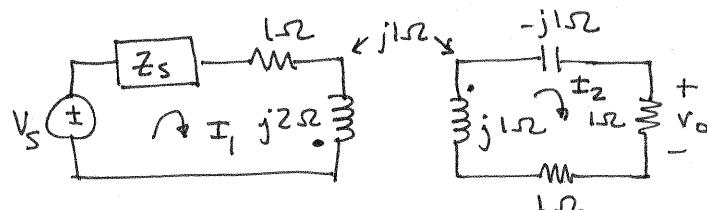
SOLUTION: Simplify through source transformation.



$$I_s = 12 \angle 0 - 12 \angle 45 = 9.18 \angle -67.5^\circ \text{ A}$$

$$z_s = 2(1-j1) / (3-j1) = 0.8 - j0.8 + j2 = 0.894 \angle -26.6^\circ \Omega$$

$$V_s = I_s z_s = 8.21 \angle -94.1^\circ \text{ V}$$



$$V_s = I_1 [z_s + 1 + j2] + j I_2$$

$$0 = j I_1 + I_2 (z)$$

$$V_o = (1) I_2$$

$$\begin{bmatrix} 1.8 + j1.6 & j1 \\ j1 & 2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 8.21 \angle -94.1^\circ \\ 0 \end{bmatrix} \Rightarrow I_2 = 1.47 \angle 141^\circ \text{ A}$$

$V_o = 1.47 \angle 141^\circ \text{ V}$

10.31 Find  $V_o$  in the network in Fig. P10.31. |cs|

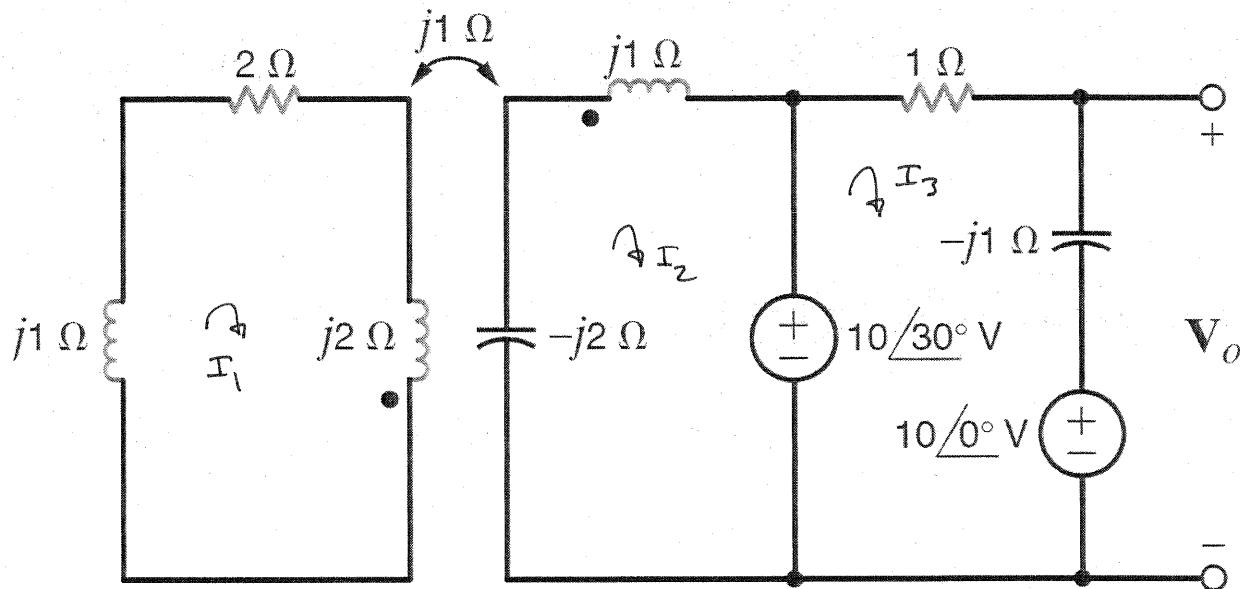


Figure P10.31

SOLUTION:

$$I_1(2+j3) - jI_2 = 0 \quad \# \quad -jI_1 + I_2(-j1) = -10 \angle 30^\circ \quad \# \quad V_o = -jI_3 + 10 \angle 0^\circ$$

$$10 \angle 30^\circ - 10 \angle 0^\circ = I_3(1-j1)$$

$$\begin{bmatrix} 2+j3 & -j1 & 0 \\ -j1 & -j1 & 0 \\ 0 & 0 & 1-j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 0 \\ -10 \angle 30^\circ \\ 10 \angle 30^\circ - 10 \angle 0^\circ \end{bmatrix} \quad I_3 = 3.66 \angle 150^\circ \text{ A}$$

$$V_o = 12.2 \angle 15.0^\circ \text{ V}$$

10.32 Find  $I_o$  in the circuit in Fig. P10.32.

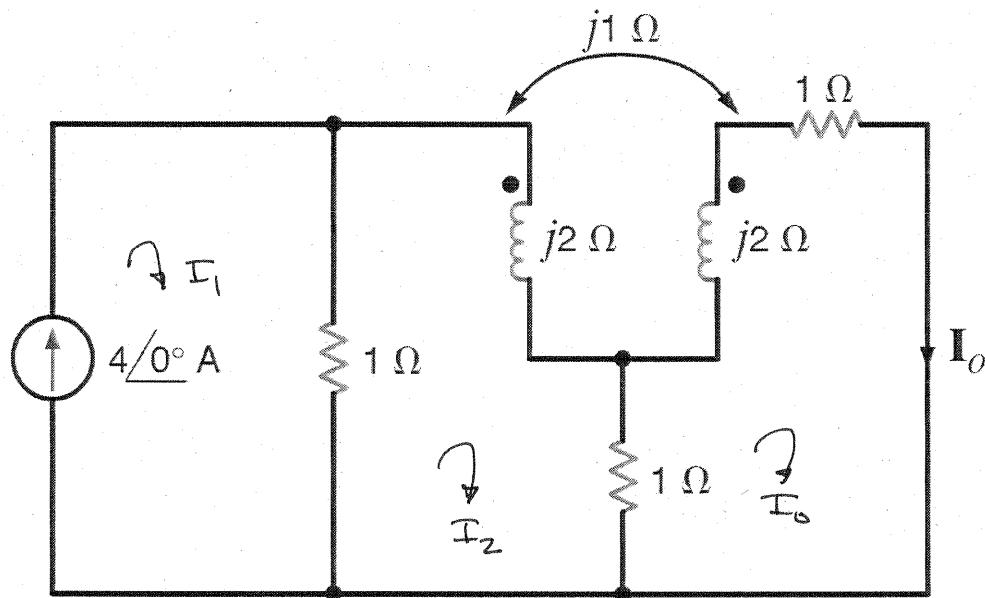


Figure P10.32

SOLUTION:

$$I_1 = 4\angle 0^\circ \neq -I_1 + I_2(z+j2) - jI_o - I_o = 0$$

$$\text{and } 0 = I_2(-1-j1) + I_o(z+j2)$$

$$\begin{bmatrix} 1 & 0 & 0 \\ -1 & 2+j2 & -1-j1 \\ 0 & -1-j1 & 2+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 4 \\ 0 \\ 0 \end{bmatrix}$$

$$\boxed{I_o = 0.667 - j0.667 \text{ A}}$$

$$\boxed{I_o = 0.943 \angle -45^\circ \text{ A}}$$

10.33 Find  $I_o$  in the circuit in Fig. P10.33.

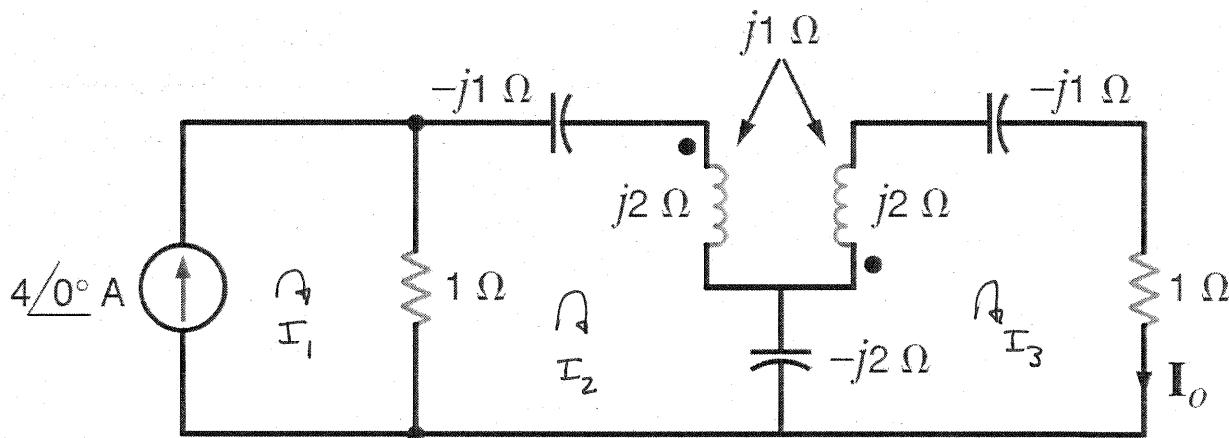


Figure P10.33

SOLUTION:

$$I_1 = 4 \angle 0^\circ \text{ A} \quad \text{and} \quad 0 = -I_1 + I_2(1-j1) + I_3(j1+j2) \quad I_3 = I_o$$

$$\text{and, } 0 = I_2(j2 + j1) + I_3(1 - j1)$$

$$\begin{bmatrix} 1 & 0 & 0 \\ -1 & 1-j1 & j3 \\ 0 & j3 & 1-j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 4 \\ 0 \\ 0 \end{bmatrix} \Rightarrow I_3 = I_o = 1.30 \angle -72.5^\circ \text{ A}$$

10.34 Find  $V_o$  in the network in Fig. P10.34.

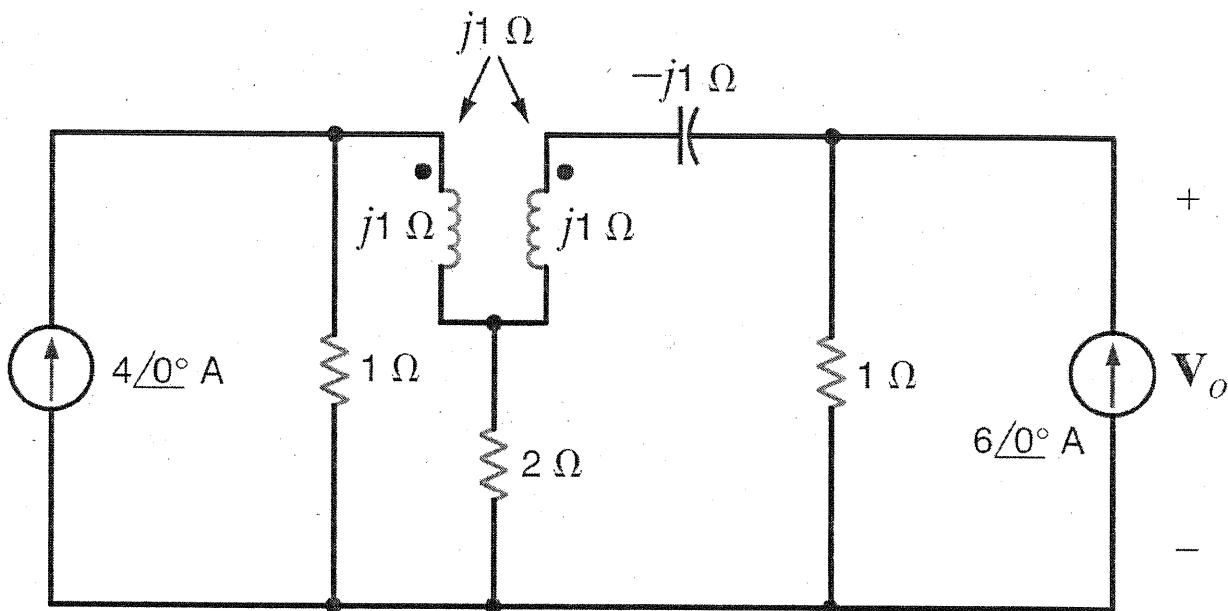
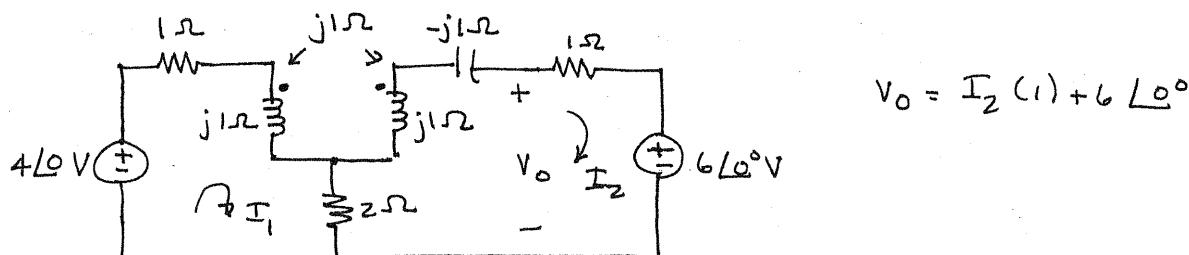


Figure P10.34

SOLUTION: Use source transformations



$$4∠0 = I_1(3+j1) - I_2(z+j1) \quad \text{and} \quad -6∠0 = -I_1(z+j1) + I_2(3)$$

$$\begin{bmatrix} 3+j1 & -z-j1 \\ -z-j1 & 3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 4 \\ -6 \end{bmatrix} \Rightarrow I_2 = 1.53 \angle -176^\circ \text{ A}$$

$$V_o = 4.47 \angle -1.51^\circ \text{ V}$$

10.35 Find  $V_o$  in the circuit in Fig. P10.35. **PSV**

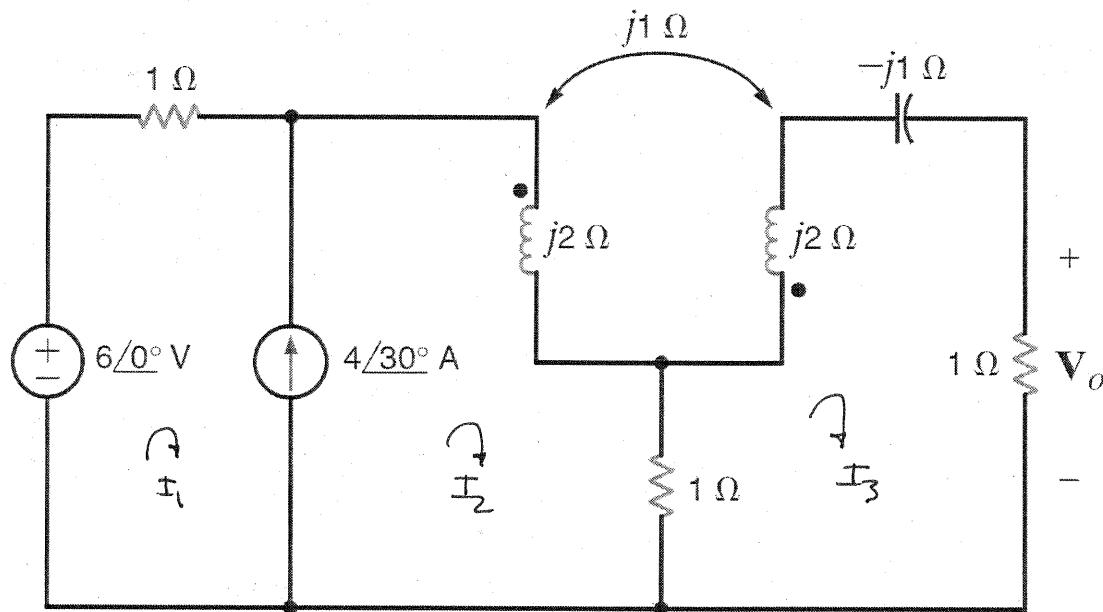


Figure P10.35

**SOLUTION:**

$$6\angle 0^\circ = I_1 + I_2(1+j2) + I_3(-1+j1) \quad \& \quad I_2(-1+j1) + I_3(2+j1) = 0$$

$$I_2 - I_1 = 4\angle 30^\circ$$

$$\begin{bmatrix} 1 & 1+j2 & -1+j1 \\ 0 & -1+j1 & 2+j1 \\ -1 & 1 & 0 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 6 \\ 0 \\ 4\angle 30^\circ \end{bmatrix} \Rightarrow I_3 = 1.66 \angle -109^\circ \text{ A}$$

$$V_o = (1) I_3$$

$$V_o = 1.66 \angle -109^\circ \text{ V}$$

10.36 Find  $V_o$  in the network in Fig. P10.36. [CS]

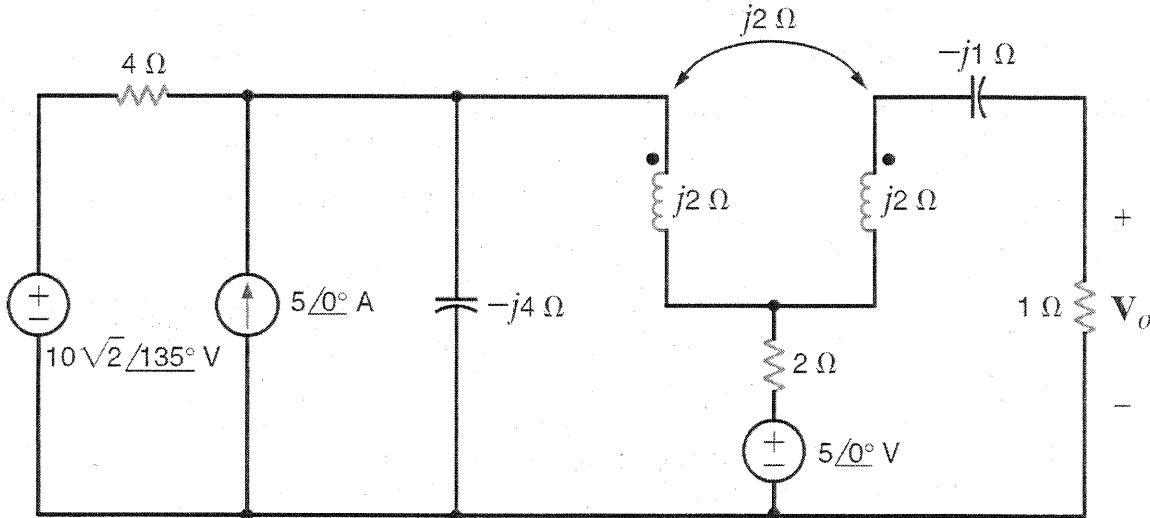
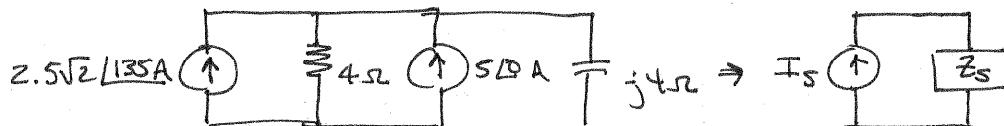


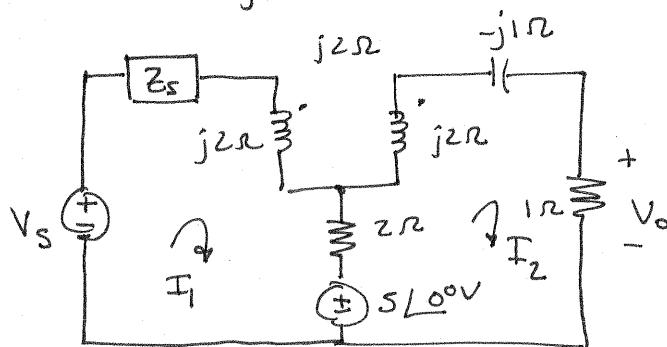
Figure P10.36

SOLUTION: Using Source transformation



$$I_s = 5 \angle 0^\circ + 2.5 \sqrt{2} \angle 135^\circ = 5 - 2.5 + j2.5 = 2.5 + j2.5 \text{ A}$$

$$Z_s = \frac{4(-j4)}{4-j4} = 2-j2 \Omega \quad V_s = I_s Z_s = 10 \angle 0^\circ \text{ V}$$



$$V_s = I_1 [Z_s + j2 + 2] + I_2 [-2 - j2] + 5 \angle 0^\circ$$

$$5 \angle 0^\circ = I_1 [-2 - j2] + I_2 [3 + j1]$$

$$\begin{bmatrix} 4 & -2 - j2 \\ -2 - j2 & 3 + j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 5 \\ 5 \end{bmatrix}$$

$$I_2 = 2.5 \angle 36.9^\circ \text{ A}$$

$$V_o = (1) I_2$$

$$V_o = 2.5 \angle 36.9^\circ \text{ V}$$

- 10.37 Determine the impedance seen by the source in the network shown in Fig. P10.37.

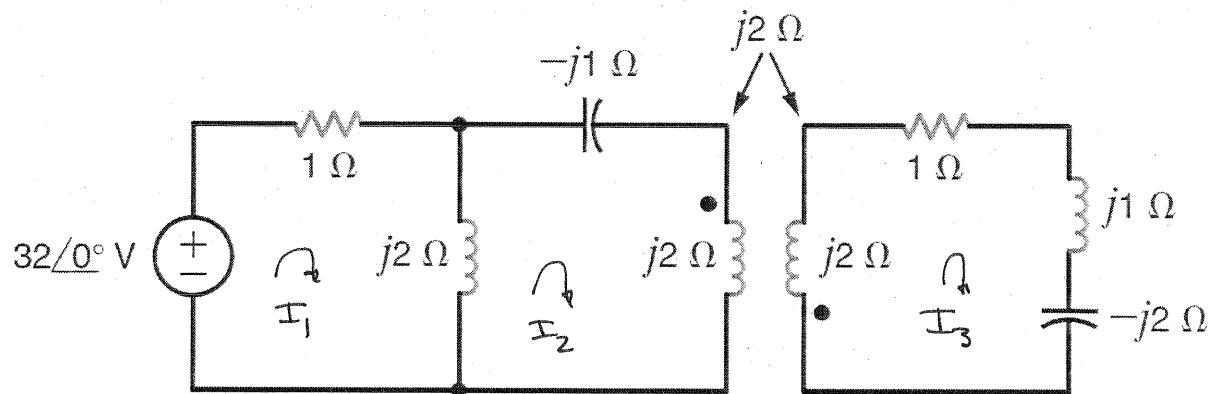


Figure P10.37

SOLUTION:

$$32\angle 0^\circ = I_1(1+j2) - j2I_2 \quad \text{and} \quad 0 = -j2I_1 + I_2[j^3] + j2I_3$$

$$\text{and} \quad j2I_2 + I_3(1+j1) = 0$$

$$Z_s = \frac{32\angle 0^\circ}{I_1}$$

$$\begin{bmatrix} 1+j2 & -j2 & 0 \\ -j2 & j^3 & j2 \\ 0 & j2 & 1+j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 32 \\ 0 \\ 0 \end{bmatrix} \Rightarrow I_1 = 11.2 \angle -24.8^\circ \text{ A}$$

$$Z_s = 2.6 + j1.25 \Omega$$

**10.38** Determine the impedance seen by the source in the network shown in Fig. P10.38.

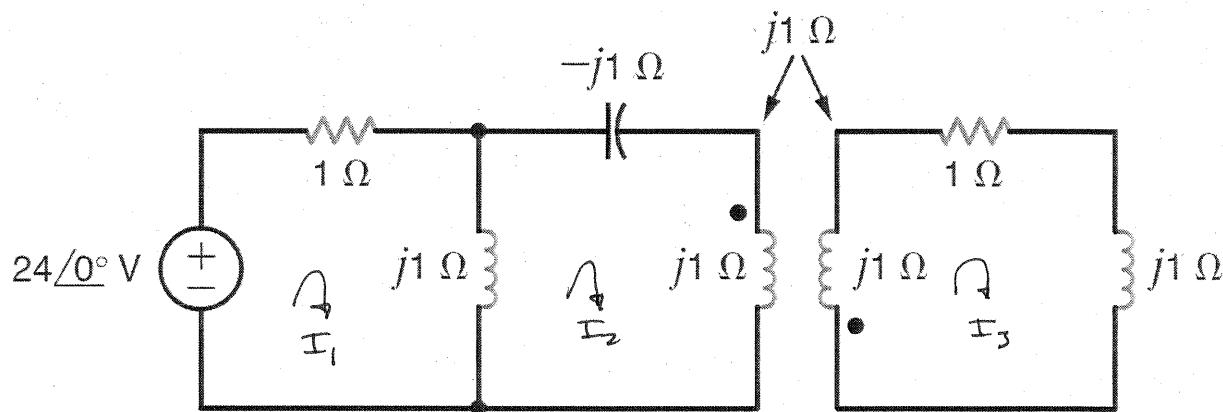


Figure P10.38

SOLUTION:

$$24\angle 0^\circ = I_1(1+j1) - jI_2 \quad & 0 = -jI_1 + I_2(j1) + jI_3$$

$$\text{and } j1I_2 + I_3(1+j2) = 0 \quad Z_S = 24\angle 0^\circ / I_1$$

$$\begin{bmatrix} 1+j1 & -j1 & 0 \\ -j1 & j1 & j1 \\ 0 & j1 & 1+j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 24 \\ 0 \\ 0 \end{bmatrix} \Rightarrow I_1 = 14.4 + j4.8 \text{ A}$$

$$Z_S = 1.5 - j0.5 \Omega$$

- 10.39 Determine the input impedance  $Z_{in}$  in the network in Fig. P10.39.

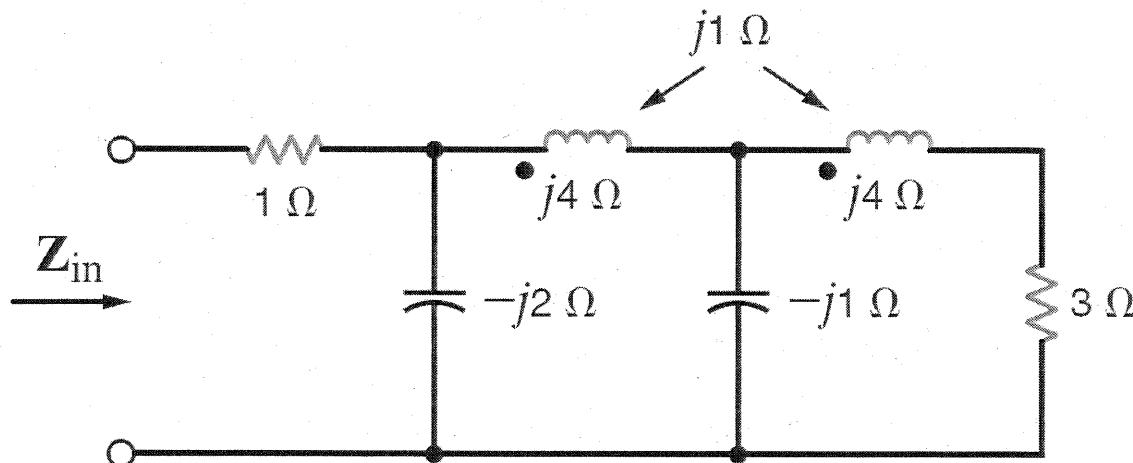
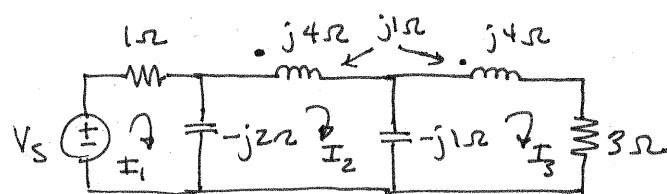


Figure P10.39

SOLUTION:



$$Z_{in} = V_s / I_1$$

let  $V_s = 1 L^0 \text{ V}$ , so that

$$Z_{in} = 1 / I_1$$

$$\left. \begin{aligned} V_s &= I_1(1-j2) + j2I_2 \\ 0 &= j2I_1 + I_2(j1) + j1I_3 + j1I_3 \\ 0 &= jI_2 + jI_2 + I_3(3+j3) \end{aligned} \right\}$$

$$\begin{bmatrix} 1-j2 & j2 & 0 \\ j2 & j1 & j2 \\ 0 & j2 & 3+j3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}$$

$$I_1 = 0.137 \angle 37.2^\circ$$

$$Z_{in} = 5.8 - j4.4 \Omega$$

- 10.40** Determine the input impedance  $Z_{in}$  of the circuit in Fig. P10.40. CS

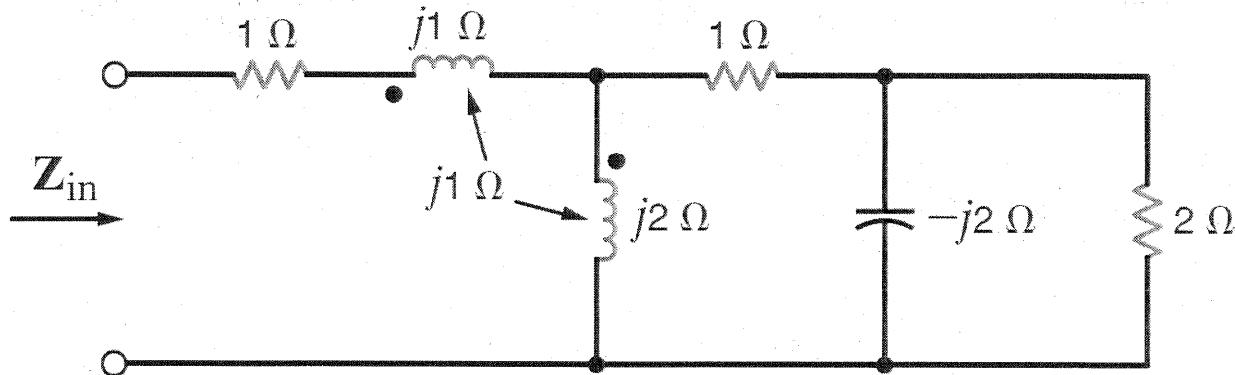
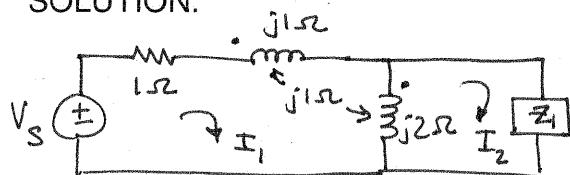


Figure P10.40

SOLUTION:



$$Z_{in} = V_s / I_1 \quad \text{let } V_s = 1 \angle 0^\circ \text{ V}$$

$$\text{so that } Z_{in} = 1 / I_1$$

$$Z_1 = 1 + \frac{(z)(-j2)}{z-j2} = z - j1\Omega$$

$$V_s = I_1 (1 + j3) - j2I_2 + jI_1 + j(I_1 - I_2)$$

$$\hookrightarrow V_s = I_1 (1 + j5) + I_2 (-j3)$$

$$0 = -j2I_1 + I_2(z_1 + j2) - j1I_1$$

$$\left[ \begin{array}{cc|c} 1+j5 & -j3 & I_1 \\ -j3 & z+j1 & I_2 \end{array} \right] \left[ \begin{array}{c} I_1 \\ I_2 \end{array} \right] = \left[ \begin{array}{c} 1 \\ 0 \end{array} \right]$$

$$I_1 = 0.1785 \angle -34.8^\circ \text{ A}$$

$$Z_{in} = 4.6 + j3.2 \Omega$$

- 10.41** Given the network shown in Fig. P10.41, determine the value of the capacitor  $C$  that will cause the impedance seen by the  $24/0^\circ$  V voltage source to be purely resistive.  $f = 60$  Hz.

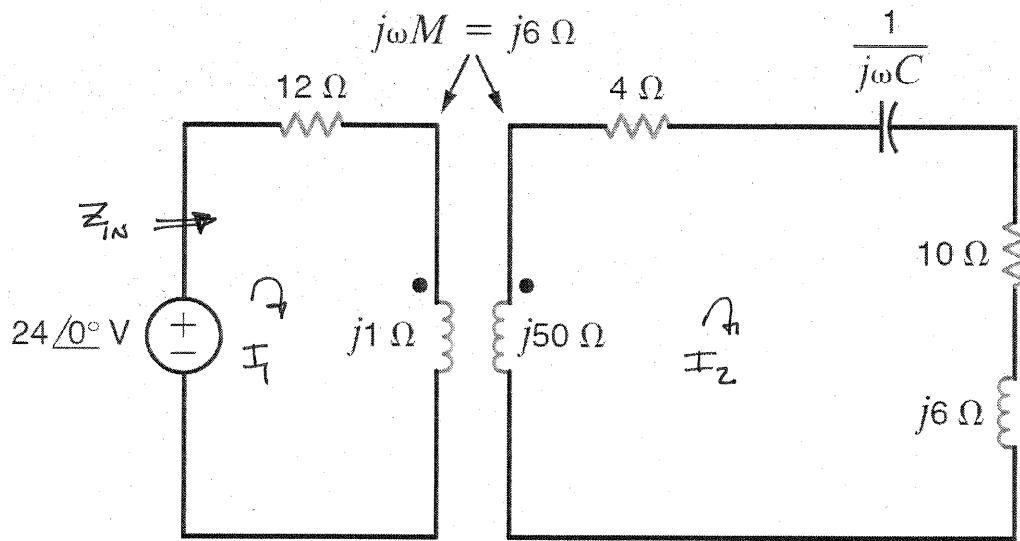


Figure P10.41

SOLUTION: For  $Z_{IN} = R_{IN} + j0$ ,  $\angle I_1 = 0^\circ$ ;  $X = \omega - 1/\omega C$

$$24/0^\circ = I_1(12 + j1) - j6I_2 \quad \& \quad -j6I_1 + I_2[14 + jX] = 0$$

$$24/0^\circ = I_1 \left[ 12 + j1 + \frac{36}{14 + jX} \right] \quad \Leftarrow I_2 = j6I_1 / (14 + jX)$$

$$\frac{24/0^\circ}{I_1} = Z_{IN} = R + j0 = 12 + j1 + \frac{36}{14 + jX} = 12 + j1 + \frac{36(14) - j36X}{14^2 + X^2}$$

$$j1 - \frac{j36X}{14^2 + X^2} = 0 \Rightarrow X^2 - 36X + 14^2 = 0 \Rightarrow X = \begin{cases} 29.3 \\ 6.69 \end{cases}$$

$$\frac{1}{\omega C} = 56 - X = \begin{cases} 49.3 \\ 26.7 \end{cases} \Rightarrow C = \begin{cases} 53.8 \mu F \\ 99 \mu F \end{cases} \text{ either will work!}$$

- 10.42** Analyze the network in Fig. P10.42 and determine whether a value of  $X_C$  can be found such that the output voltage is equal to twice the input voltage.

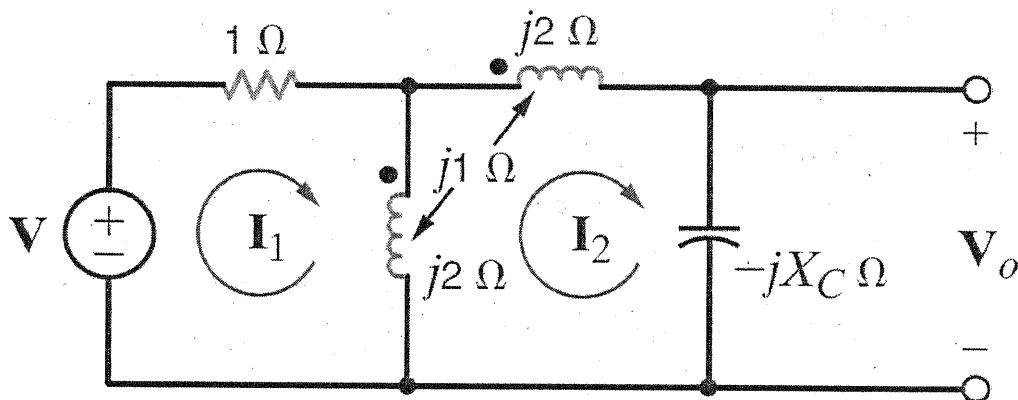


Figure P10.42

SOLUTION:

$$V = I_1(1+j2) - j2I_2 + jI_2 \Rightarrow V = I_1(1+j2) - jI_2$$

$$0 = -j2I_1 + I_2(j(4-X_C)) - jI_2 + j(I_1 - I_2) \Rightarrow 0 = -jI_1 + jI_2(z-X_C)$$

$$\text{or, } I_1 = I_2(z-X_C)$$

$$\text{now, } V = I_2[(1+j2)(z-X_C) - j1] \Rightarrow V = I_2[z-X_C + j(3-2X_C)]$$

$$\text{But, } V_o = -jX_C I_2$$

$$\text{So, } \frac{V_o}{V} = \frac{-jX_C}{z-X_C + j(3-2X_C)} = z + j0$$

$$\text{requires } z-X_C = 0 \Rightarrow X_C = z \quad \text{and} \quad \frac{V_o}{V} = \frac{-jz}{0+j(3-4)} = z + j0 \quad \checkmark$$

$$\boxed{X_C = z \Omega}$$

- 10.43** Two coils in a network are positioned such that there is 100% coupling between them. If the inductance of one coil is 10 mH and the mutual inductance is 6 mH, compute the inductance of the other coil. **CS**

---

SOLUTION:

$$k = 1 = \frac{M}{\sqrt{L_1 L_2}} = \frac{6 \times 10^{-3}}{\sqrt{10^{-2} L_2}}$$

$L_2 = 3.6 \text{ mH}$

- 10.44 The currents in the network in Fig. P10.44 are known to be  $i_1(t) = 10 \cos(377t - 30^\circ)$  mA and  $i_2(t) = 20 \cos(377t - 45^\circ)$  mA. The inductances are  $L_1 = 2$  H,  $L_2 = 2$  H, and  $k = 0.8$ . Determine  $v_1(t)$  and  $v_2(t)$ .

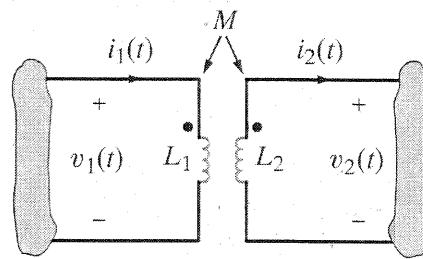
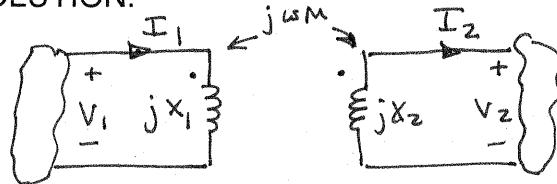


Figure P10.44

SOLUTION:



$$X_1 = \omega L_1 = 754 \Omega$$

$$X_2 = \omega L_2 = 754 \Omega$$

$$M = k\sqrt{L_1 L_2} = 1.6 \text{ H} \quad \omega M = 603 \Omega$$

$$V_1 = I_1(jX_1) - j\omega M I_2$$

$$I_1 = 10 \angle -30^\circ \text{ mA}$$

$$-V_2 = -j\omega M I_1 + I_2 jX_2$$

$$I_2 = 20 \angle -45^\circ \text{ mA}$$

$$V_1 = 5.16 \angle -157^\circ \text{ V}$$

$$V_2 = 9.39 \angle -145^\circ \text{ V}$$

$$v_1(t) = 5.16 \cos(377t - 157^\circ) \text{ V}$$

$$v_2(t) = 9.39 \cos(377t - 145^\circ) \text{ V}$$

- 10.45 Determine the energy stored in the coupled inductors in the circuit in P10.44 at  $t = 1 \text{ ms}$ .

---

SOLUTION:

$$w(t) = \frac{1}{2} L_1 i_1^2(t) + \frac{1}{2} L_2 i_2^2(t) - m i_1(t) i_2(t)$$

$$i_1(t) = 10 \cos(377t - 30^\circ) \text{ mA} \quad i_2(t) = 20 \cos(377t - 45^\circ) \text{ mA}$$

$$L_1 = 2 \text{ H} \quad L_2 = 2 \text{ H} \quad k = 0.8 \Rightarrow M = 1.6 \text{ H}$$

$$\text{let } t_1 = 1 \text{ ms}$$

$$i_1(t_1) = 9.89 \text{ mA} \quad i_2(t_1) = 18.4 \text{ mA}$$

$$w(t_1) = 145 \mu\text{J}$$

- 10.46 The currents in the magnetically-coupled inductors shown in Fig. P10.46 are known to be  $i_1(t) = 8 \cos(377t - 20^\circ)$  mA and  $i_2(t) = 4 \cos(377t - 50^\circ)$  mA. The inductor values are  $L_1 = 2$  H,  $L_2 = 1$  H, and  $k = 0.6$ . Determine  $v_1(t)$  and  $v_2(t)$ .

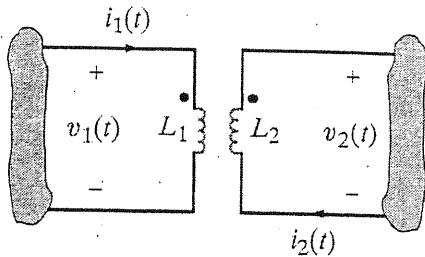


Figure P10.46

$$\text{SOLUTION: } M = k\sqrt{L_1 L_2} \Rightarrow M = 0.849 \text{ H}$$

$$v_1(t) = L_1 \frac{di_1}{dt} - M \frac{di_2}{dt} = \frac{-2(8)(377)}{1000} \sin(377t - 20^\circ) + \frac{M(4)(377)}{1000} \sin(377t - 50^\circ)$$

$$M = k\sqrt{L_1 L_2} \Rightarrow M = 0.849 \text{ H} \quad I_1 = 8 \angle -20^\circ \text{ mA} \quad I_2 = 4 \angle -50^\circ \text{ mA}$$

$$V_1 = j\omega L_1 I_1 - j\omega M I_2 \quad \omega = 377 \text{ rad/s} \quad \omega M = 320 \Omega$$

$$-V_2 = -j\omega M I_1 + j\omega L_2 I_2$$

$$V_1 = 4.96 \angle 77.4^\circ \text{ V} \quad V_2 = 5.12 \angle 70^\circ \text{ V}$$

$$v_1(t) = 4.96 \cos(377t + 77.4^\circ) \text{ V}$$

$$v_2(t) = 1.46 \cos(377t + 101^\circ) \text{ V}$$

**10.47** Determine the energy stored in the coupled inductors in Problem 10.46 at  $t = 1 \text{ ms}$ . **CS**

SOLUTION:  $W(t) = \frac{1}{2} L_1 i_1^2(t) + \frac{1}{2} L_2 i_2^2(t) + M i_1(t) i_2(t)$

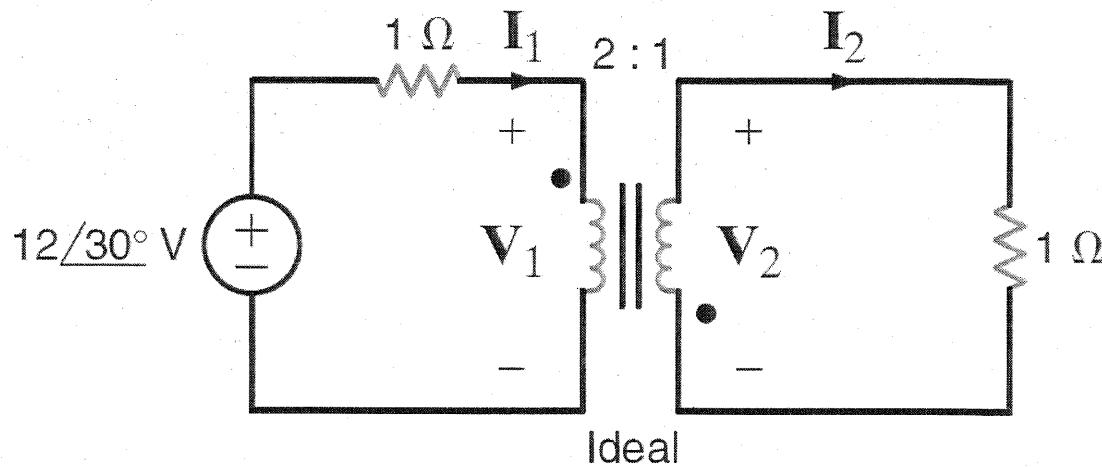
$$i_1(t) = 8 \cos(377t - 20^\circ) \text{ mA} \quad i_2(t) = 4 \cos(377t - 50^\circ) \text{ mA}$$

$$L_1 = 2 \text{ H} \quad L_2 = 1 \text{ H} \quad k = 0.6 \Rightarrow M = 0.849 \text{ H}$$

$$\text{let } t_1 = 1 \text{ ms}, \quad i_1(t_1) = 8.00 \text{ mA} \quad i_2(t_1) = 3.52 \text{ mA}$$

$W(t_1) = 94.1 \mu\text{J}$

**10.48** Determine  $I_1$ ,  $I_2$ ,  $V_1$ , and  $V_2$  in the network in Fig. P10.48.



**Figure P10.48**

**SOLUTION:**

$$12 \angle 30^\circ = (1)I_1 + V_1 \quad \& \quad V_2 = (1)I_2 \quad \& \quad -V_1 = 2V_2 \quad \& \quad 2I_1 = -I_2$$

$$12 \angle 30^\circ = I_1 - 2V_2 = I_1 - 2I_2 = I_1 + 4I_1 = 5I_1$$

$$I_1 = \frac{12 \angle 30^\circ}{5} = 2.4 \angle 30^\circ \text{ A} = I_1$$

$$I_2 = 4.8 \angle -150^\circ \text{ A}$$

$$V_2 = 4.8 \angle -150^\circ \text{ V}$$

$$V_1 = 9.6 \angle 30^\circ \text{ V}$$

- 10.49 Find all currents and voltages in the network in Fig. P10.49.

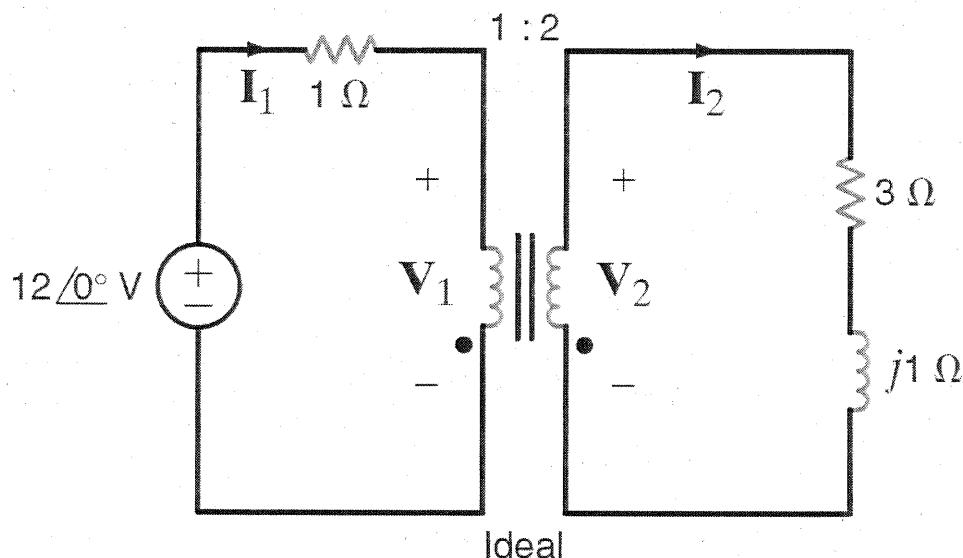
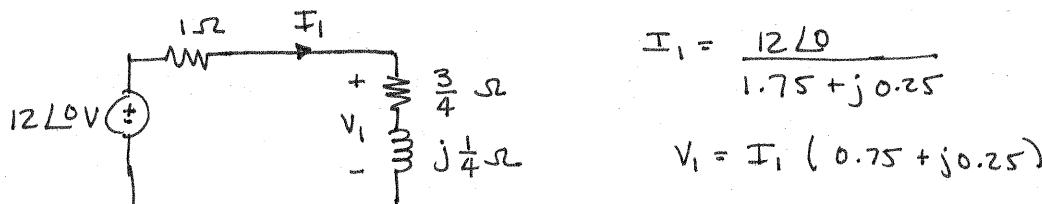


Figure P10.49

SOLUTION:  $n = 2$        $n^2 = 4$

$$12\angle 0^\circ = (1)I_1 + V_1 \quad \# \quad V_2 = I_2(3+j1) \quad \# \quad V_2 = 2V_1 \quad \# \quad I_1 = 2I_2$$



$$I_1 = \frac{12\angle 0^\circ}{1.75 + j0.25}$$

$$V_1 = I_1 (0.75 + j0.25)$$

$$I_1 = 6.79 \angle -8.13^\circ \text{ A} \quad I_2 = 3.39 \angle -8.13^\circ \text{ A}$$

$$V_1 = 5.37 \angle 10.3^\circ \text{ V} \quad V_2 = 10.74 \angle 10.3^\circ \text{ V}$$

**10.50** Determine  $V_o$  in the circuit in Fig. P10.50.

**PSV**

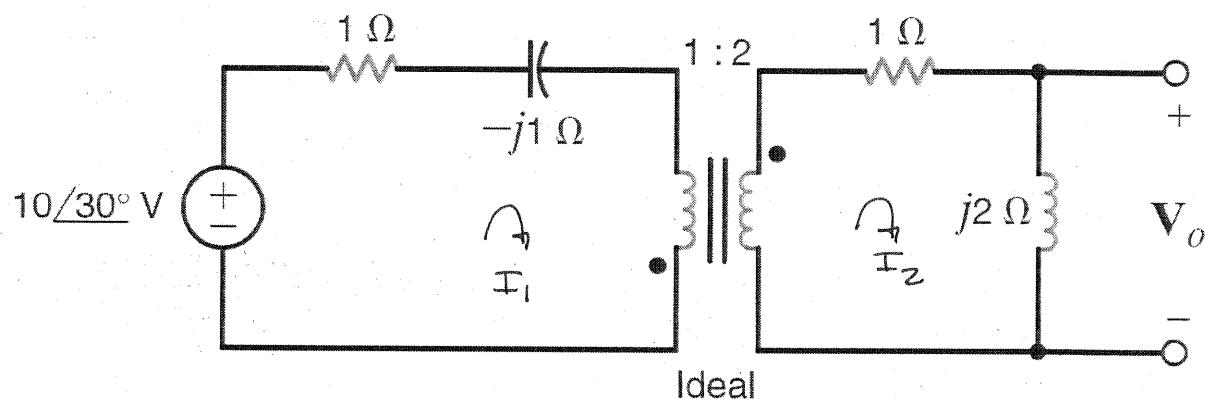
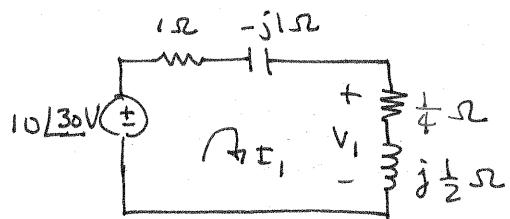


Figure P10.50

SOLUTION:  $n = 2 \quad n^2 = 4$

Use impedance reflection,



$$I_2 = -I_1 / 2 \quad V_o = j2 I_2$$

$$I_1 = \frac{10 \angle 30^\circ}{1.25 - j0.5} = 7.43 \angle 51.8^\circ A$$

$$I_2 = 3.72 \angle -128^\circ A$$

$$V_o = 7.43 \angle -38.2^\circ V$$

10.51 Determine  $V_o$  in the circuit in Fig. P10.51.

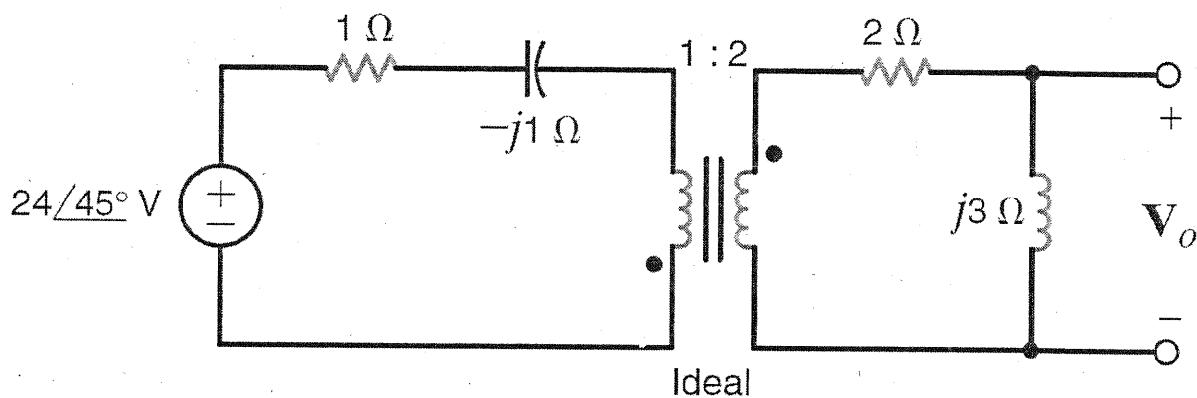
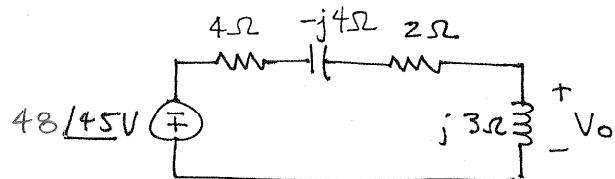


Figure P10.51

SOLUTION:  $n = 2 \quad n^2 = 4$  Using reflection through transformer



$$V_o = -\frac{48 \angle 45^\circ (j3)}{6 - j4 + j3} = -\frac{48 \angle 45^\circ (j3)}{6 - j1}$$

$$V_o = 23.7 \angle 35.5^\circ V$$

- 10.52 Determine  $I_1$ ,  $I_2$ ,  $V_1$ , and  $V_2$  in the network in Fig. P10.52. 

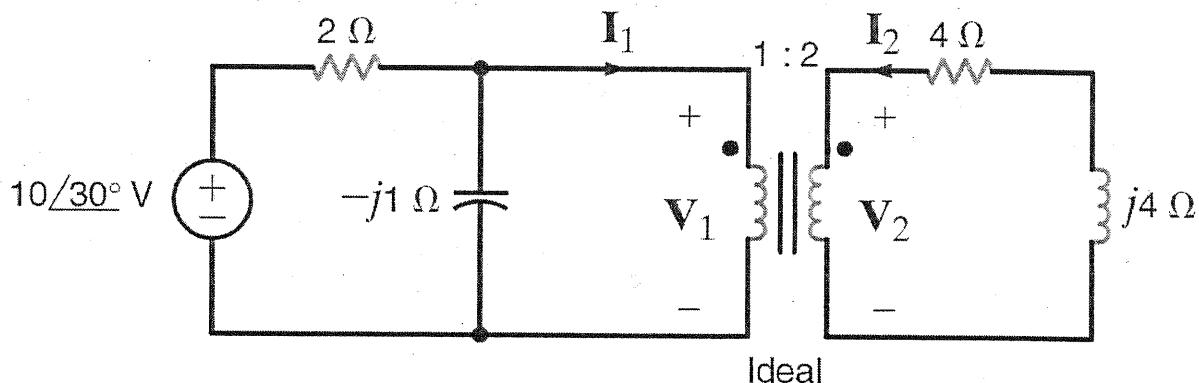
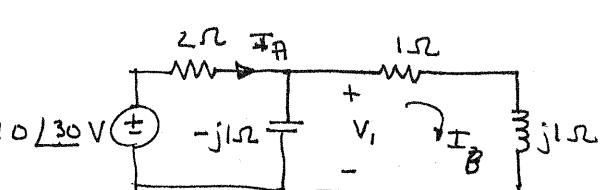


Figure P10.52

SOLUTION:  $n = 2 \quad n^2 = 4$



$$10\angle 30^\circ = I_A(2 - j1) + jI_B$$

$$0 = jI_A + I_B(1)$$

$$I_A = jI_B$$

$$10\angle 30^\circ = I_B(1 + j2 + j1) = I_B(1 + j3)$$

$$V_1 = I_B(1 + j1)$$

$$V_1 = 4.47 \angle 3.4^\circ V$$

$$V_2 = nV_1 = 8.93 \angle 3.4^\circ V$$

$$I_2 = -I_1/n = 1.58 \angle 138^\circ A$$

- 10.53 Determine  $I_1$ ,  $I_2$ ,  $V_1$ , and  $V_2$  in the network in Fig. P10.53.

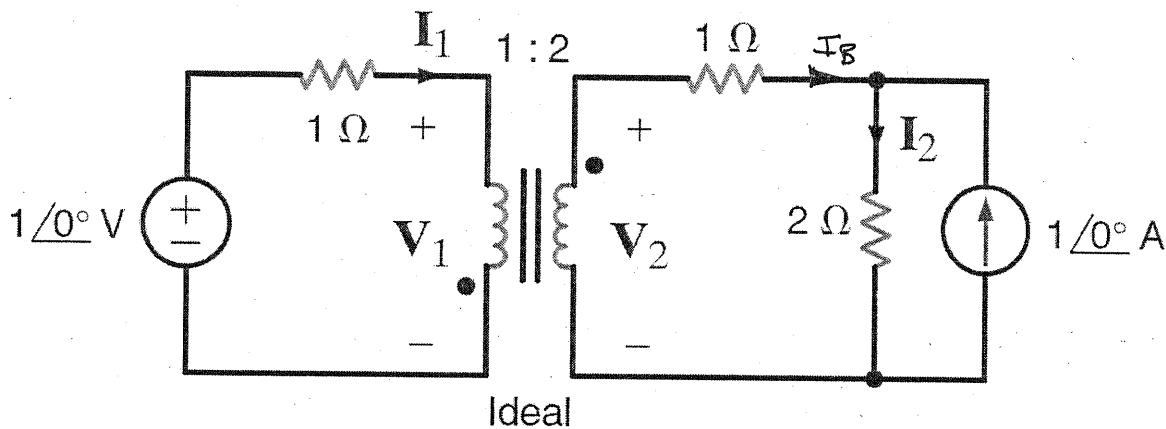
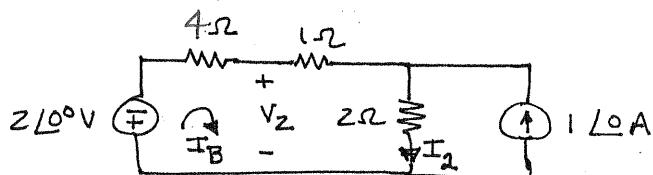


Figure P10.53

$$\text{SOLUTION: } n = z = -\frac{V_2}{V_1} = -\frac{I_1}{I_B}$$



$$-2∠0 = I_B(4 + 1 + 2) + 2(1∠0)$$

$$I_B = 0.571 \angle 180^\circ \text{ A}$$

$$I_2 = I_B + 1∠0$$

$$I_2 = 0.429 \angle 0^\circ \text{ A}$$

$$V_2 = 2I_2 + I_B(1)$$

$$V_2 = 0.287 \angle 0^\circ \text{ V}$$

$$I_1 = -n I_B$$

$$I_1 = 1.14 \angle 0^\circ \text{ A}$$

$$V_1 = -V_2/n$$

$$V_1 = 0.143 \angle 180^\circ \text{ V}$$

10.54 Find  $I$  in the network in Fig. P10.54.

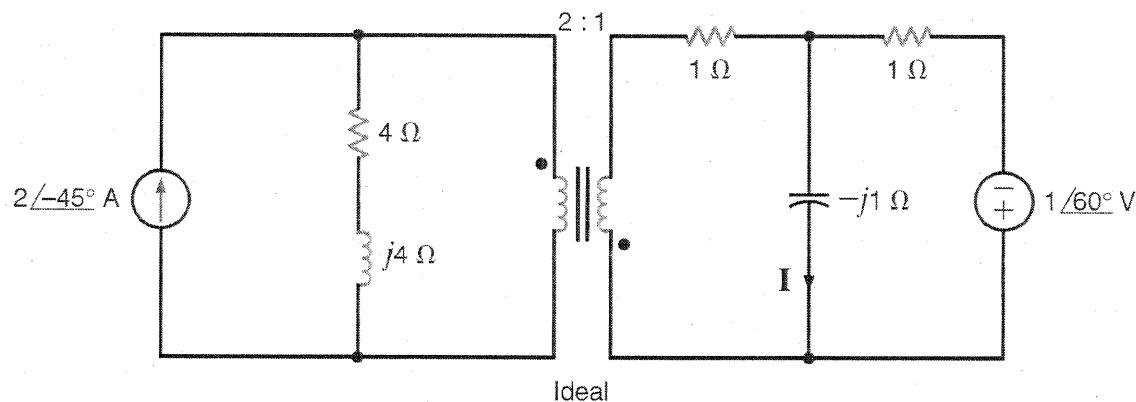
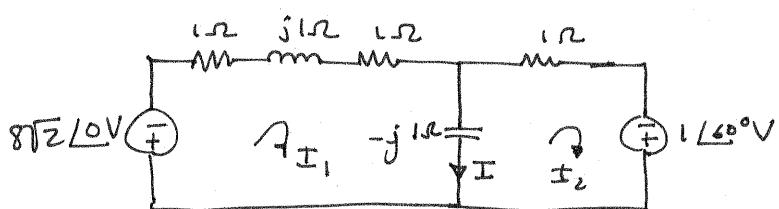
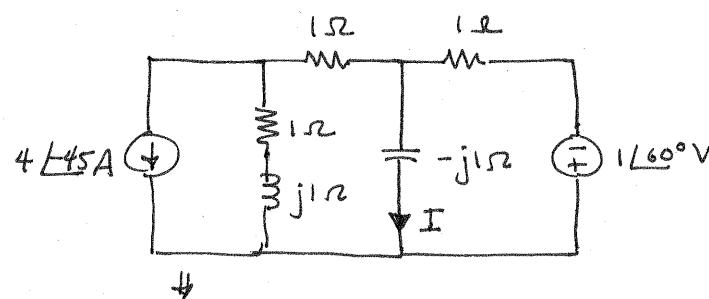


Figure P10.54

SOLUTION:  $n = 1/2$



$$\begin{aligned} -8\sqrt{2}\angle 0 &= I_1(2) + j1I_2 \\ 1\angle 60 &= j1I_1 + (1-j1)I_2 \end{aligned} \quad \left\{ \begin{bmatrix} 2 & j1 \\ j1 & 1-j1 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} -8\sqrt{2}\angle 0 \\ 1\angle 60 \end{bmatrix} \right.$$

$$I_1 = 4.17 \angle 168^\circ \text{ A} \quad I_2 = 3.63 \angle 119^\circ \text{ A}$$

$$I = I_1 - I_2 \quad \boxed{I = 3.23 \angle -135^\circ \text{ A}}$$

- 10.55 Determine  $\mathbf{I}_1$ ,  $\mathbf{I}_2$ ,  $\mathbf{V}_1$ , and  $\mathbf{V}_2$  in the network in Fig. P10.55.

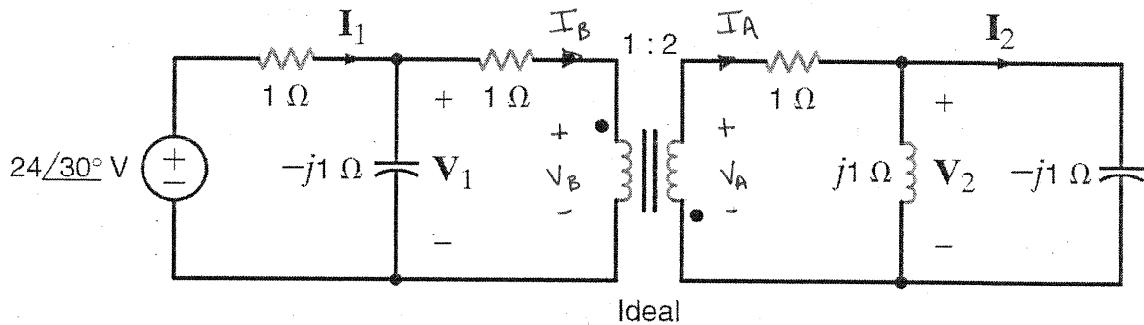


Figure P10.55

SOLUTION:  $n = 2$

$$\begin{aligned}
 & \text{Circuit diagram: } \\
 & \text{Left: } 24\angle 30^\circ \text{ V source, } 1\Omega \text{ resistor, } V_1 \text{ node.} \\
 & \text{Between } V_1 \text{ and } V_B: 1\Omega \text{ resistor, } -j1\Omega \text{ inductor, } I_1 \text{ flowing right.} \\
 & \text{Between } V_B \text{ and } V_2: \frac{1}{2}\Omega \text{ resistor, } j1\Omega \text{ inductor, } I_B \text{ flowing right.} \\
 & \text{Between } V_2 \text{ and right: } 1\Omega \text{ resistor, } -j1\Omega \text{ capacitor, } I_2 \text{ flowing right.} \\
 & Z_1 = \frac{(j1)(-j1)}{j1-j1} = \infty \\
 & I_2 = \frac{V_2}{-j1}
 \end{aligned}$$

$$24\angle 30^\circ = I_1(1-j1) + j I_B \quad * \quad I_B = 0$$

$$V_1 = (I_1 - I_B)(-j1) = 10\sqrt{2} \angle -15^\circ$$

$$V_B = -\frac{V_2}{2} = V_1 \Rightarrow V_2 = 34.0 \angle 165^\circ \text{ V}$$

$$\begin{cases} V_1 = 17.0 \angle -15^\circ \text{ V} \\ I_1 = 17.0 \angle 75^\circ \text{ A} \end{cases}$$

$$I_2 = \frac{V_2}{-j1} \Rightarrow I_2 = 34.0 \angle -105^\circ \text{ A}$$

10.56 Find the current  $\mathbf{I}$  in the network in Fig. P10.56.

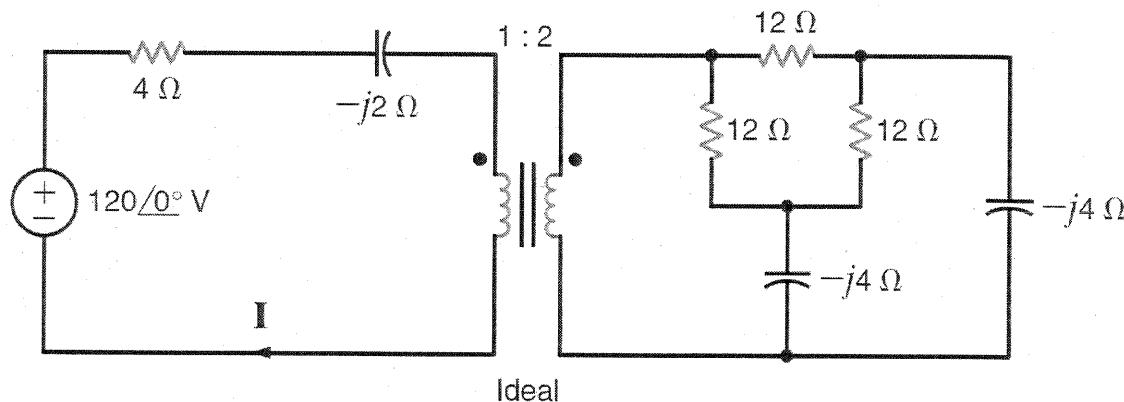
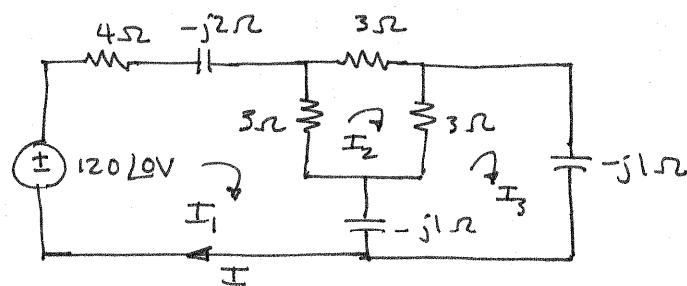


Figure P10.56

SOLUTION:  $\eta = 2$



$$120\angle 0^\circ = I_1(7-j3) - 3I_2 + jI_3$$

$$0 = -3I_1 + 9I_2 - 3I_3$$

$$0 = jI_1 - 3I_2 + I_3(3-j2)$$

$$\begin{bmatrix} 7-j3 & -3 & j1 \\ -3 & 9 & -3 \\ j1 & -3 & 3-j2 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} 120 \\ 0 \\ 0 \end{bmatrix}$$

$$I = I_1$$

$$I_1 = 19.9 \angle 24.4^\circ \text{ A}$$

$$I_0 = 19.9 \angle 24.4^\circ \text{ A}$$

- 10.57 Determine  $I_1$ ,  $I_2$ ,  $V_1$ , and  $V_2$  in the network in Fig. P10.57.

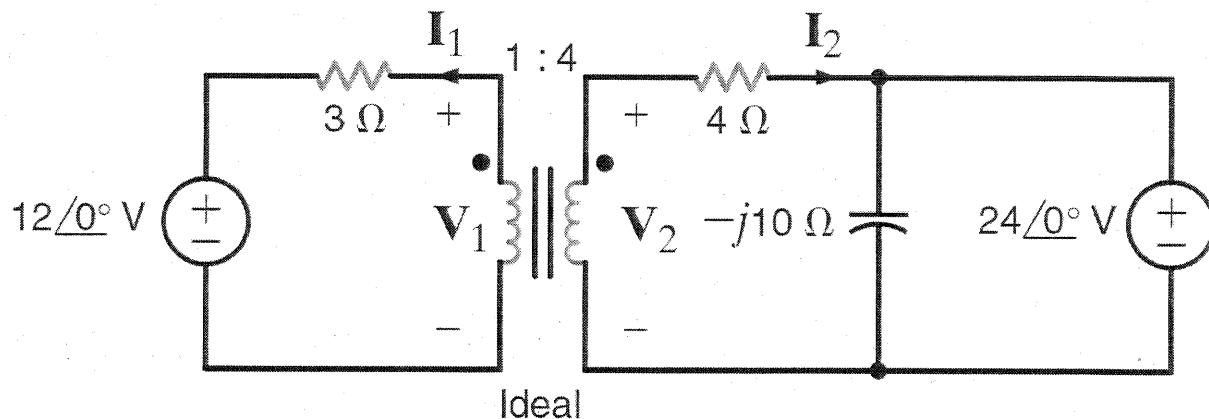
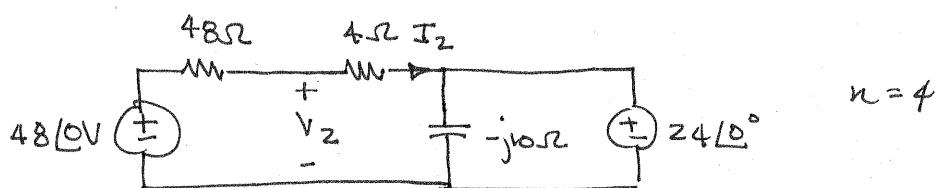


Figure P10.57

SOLUTION:



$$I_2 = \frac{48\angle 0^\circ - 24\angle 0^\circ}{48 + 4} = \frac{24\angle 0^\circ}{52} = 0.46\angle 0^\circ \text{ A}$$

$$V_2 = 48\angle 0^\circ - 48I_2 \Rightarrow V_2 = 25.9\angle 0^\circ \text{ V}$$

$$V_1 = V_2/n = 6.48\angle 0^\circ \text{ V}$$

$$I_1 = -I_2 n = -1.84\angle 0^\circ \text{ A}$$

$I_1 = -1.84\angle 0^\circ \text{ A}$	$V_1 = 6.48\angle 0^\circ \text{ V}$
$I_2 = 0.46\angle 0^\circ \text{ A}$	$V_2 = 25.9\angle 0^\circ \text{ V}$

10.58 Find  $V_o$  in the network in Fig. P10.58.

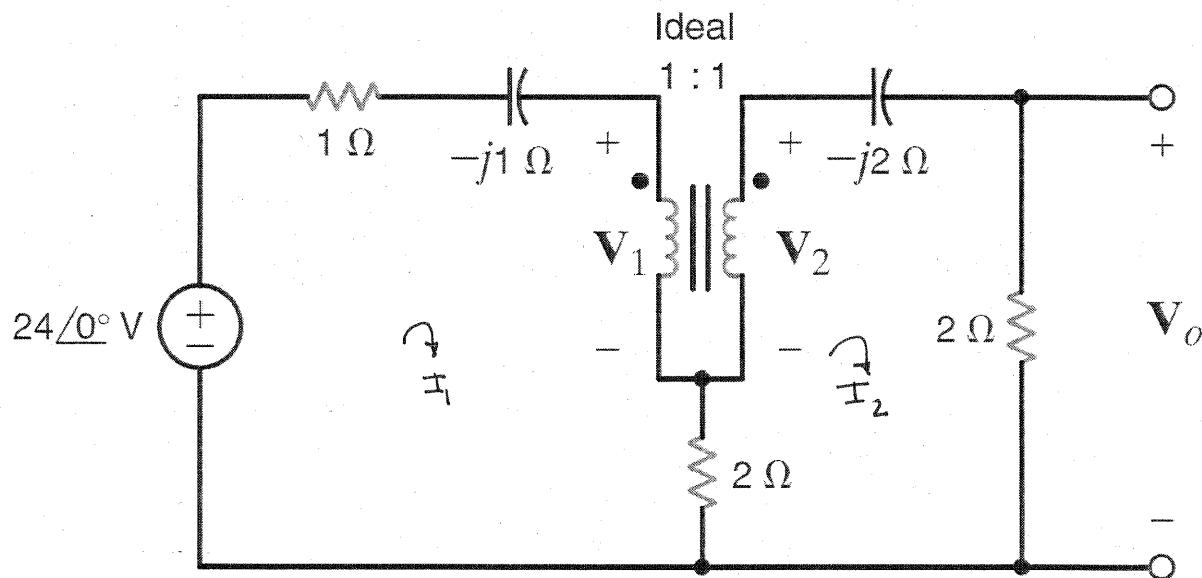


Figure P10.58

$$\text{SOLUTION: } n = 1 \quad I_1 = I_2 \quad V_1 = V_2$$

$$24\angle 0^\circ = I_1(3-j1) - zI_2 + V_1 \quad \left. \begin{array}{l} \\ \end{array} \right\} \Rightarrow 24\angle 0^\circ = I_1(1-j1) + V_1$$

$$V_2 = -zI_1 + I_2(4-j2) \quad \left. \begin{array}{l} \\ \end{array} \right\} \quad V_1 = I_1(z-j2)$$

$$24\angle 0^\circ = I_1(1-j1) + I_1(z-j2) = I_1(3-j3) \Rightarrow I_1 = \frac{24\angle 0^\circ}{3-j3} = 5.66\angle 45^\circ A$$

$$I_2 = I_1 = 5.66\angle 45^\circ A$$

$$V_o = zI_2 \quad \boxed{V_o = 11.32\angle 45^\circ V}$$

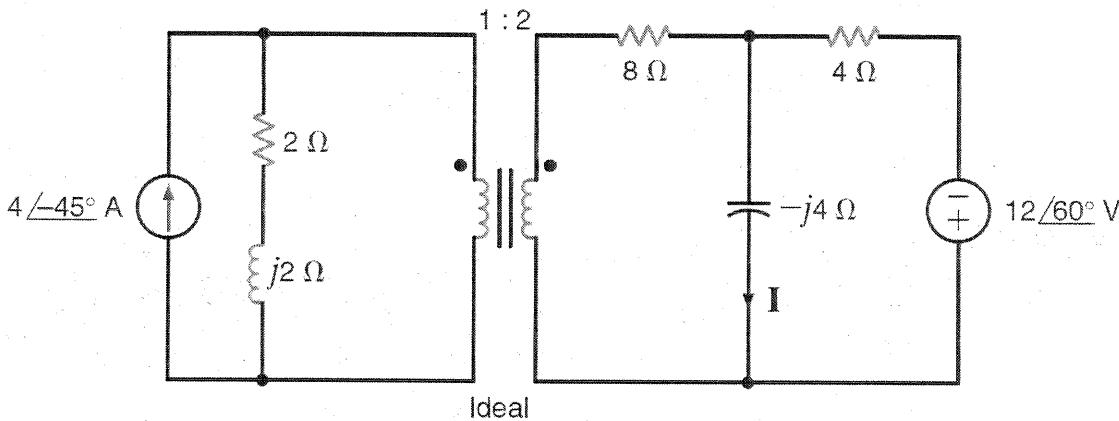
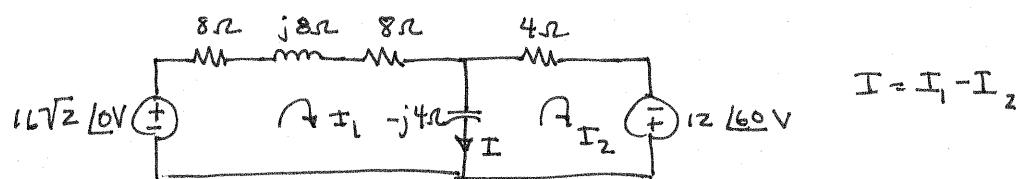
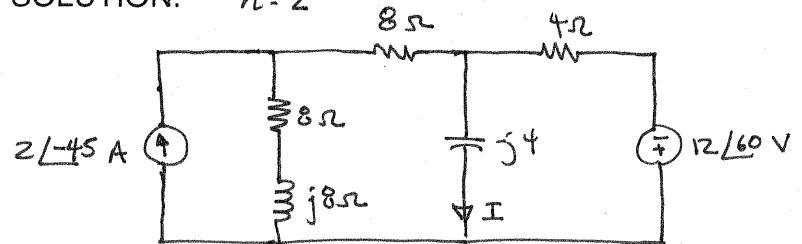
10.59 Find the current  $\mathbf{I}$  in the network in Fig. P10.59. 

Figure P10.59

SOLUTION:  $n = 2$



$$\begin{aligned} 16\sqrt{2}\angle 0^\circ &= I_1(16 + j4) + j4I_2 \\ 12\angle 60^\circ &= j4I_1 + I_2(4 - j4) \end{aligned} \quad \left. \begin{bmatrix} 16+j4 & j4 \\ j4 & 4-j4 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 16\sqrt{2} \\ 12\angle 60^\circ \end{bmatrix} \right.$$

$$I_1 = 1.63 \angle -14.3^\circ A \quad I_2 = 1.06 \angle 88.0^\circ A$$

$$I = 2.12 \angle -43.5^\circ A$$

10.60 Find the voltage  $V_o$  in the network in Fig. P10.60. **PSV**

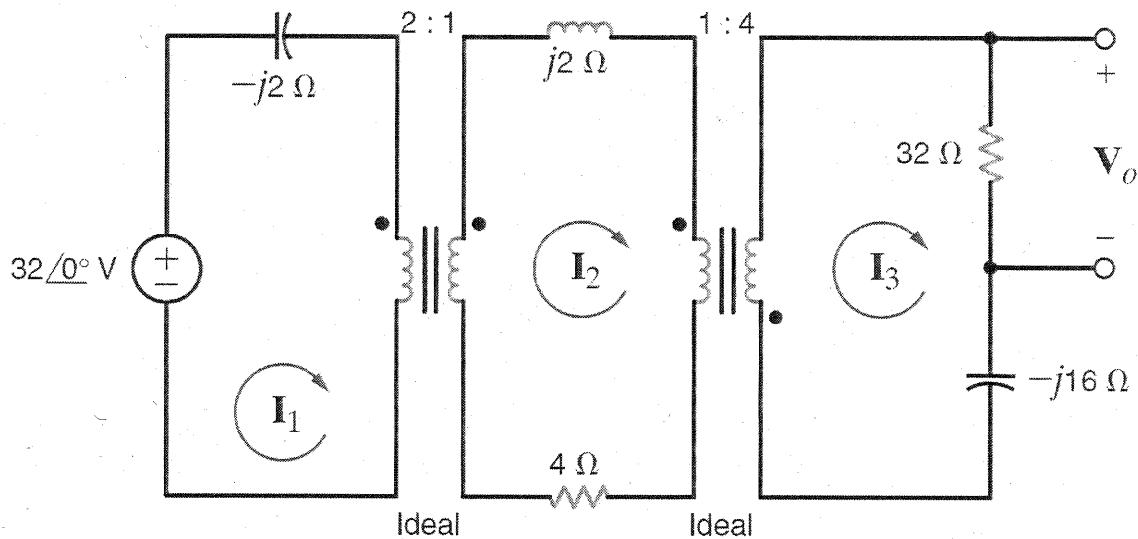
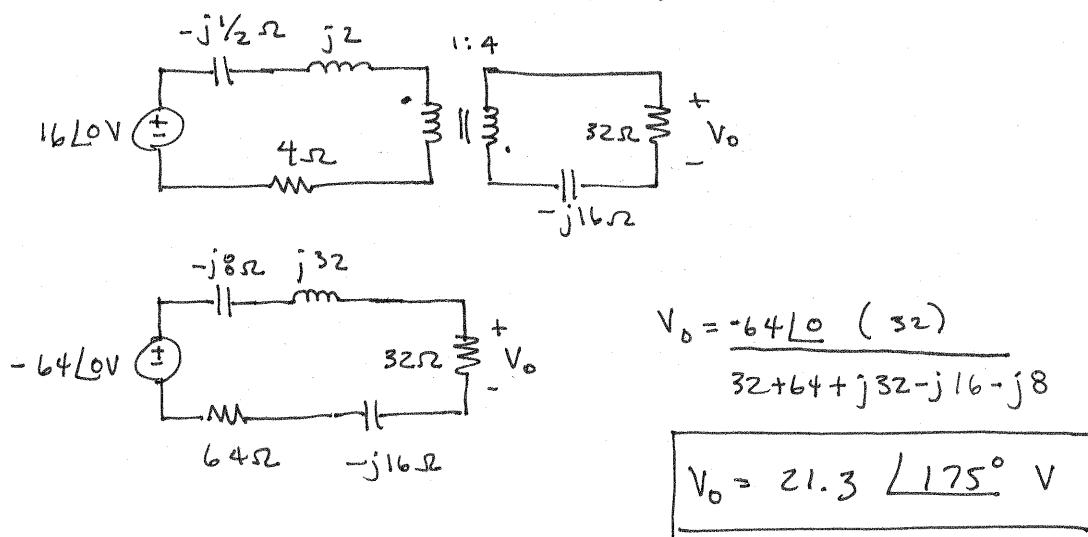


Figure P10.60

SOLUTION:  $n_1 = \frac{1}{2}$        $n_2 = 4$



10.61 Find  $\mathbf{V}_o$  in the circuit in Fig. P10.61.

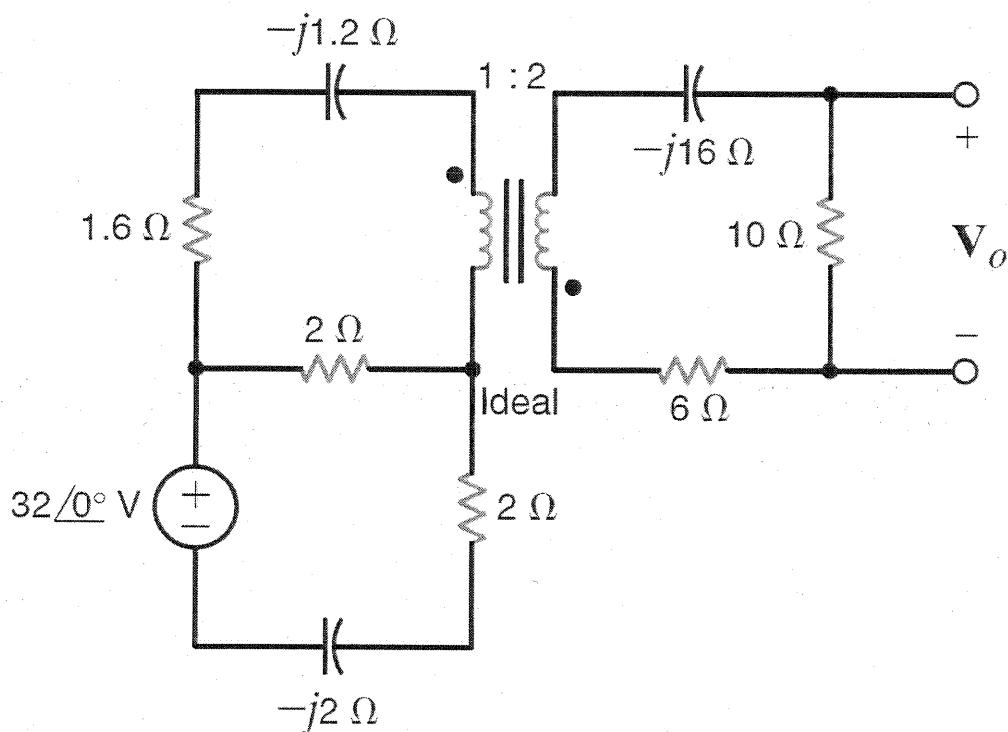
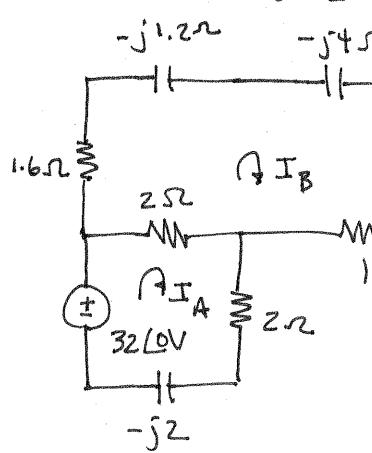


Figure P10.61

SOLUTION:  $n = 2$



$$32\angle 0^\circ = I_A (4-j2) - 2I_B$$

$$0 = -2I_A + I_B (7.6-j5.2)$$

$$\frac{V_o}{2} = -I_B (2.5) \Rightarrow V_o = -5I_B$$

$$\begin{bmatrix} 4-j2 & -2 \\ -2 & 7.6-j5.2 \end{bmatrix} \begin{bmatrix} I_A \\ I_B \end{bmatrix} = \begin{bmatrix} 32 \\ 0 \end{bmatrix}$$

$$I_B = 1.62 \angle 66.0^\circ \text{ A}$$

$$V_o = 8.12 \angle -114^\circ \text{ V}$$

10.62 Find  $V_o$  in the network in Fig. P10.62.

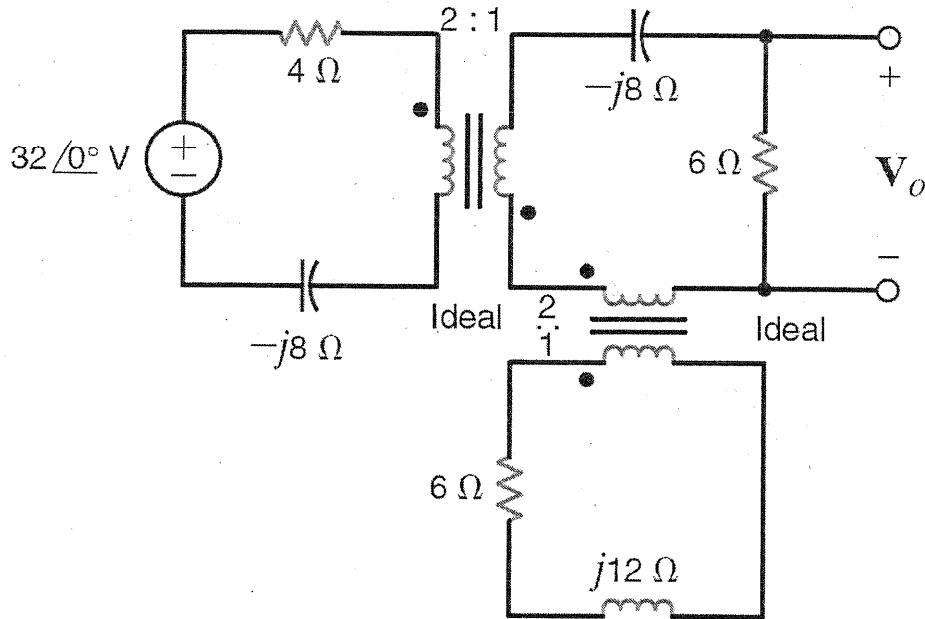
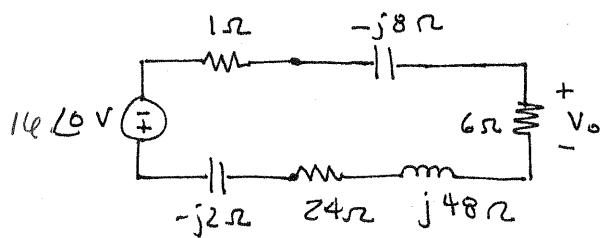


Figure P10.62

SOLUTION:  $n_1 = \frac{1}{2}$      $n_2 = 2$



$$V_o = \frac{-16\angle 0^\circ (6)}{1 + 6 + 24 + j(48 - 8 - 2)}$$

$$V_o = 1.96 \angle 129^\circ V$$

10.63 Find  $V_o$  in the circuit in Fig. P10.63. | cs

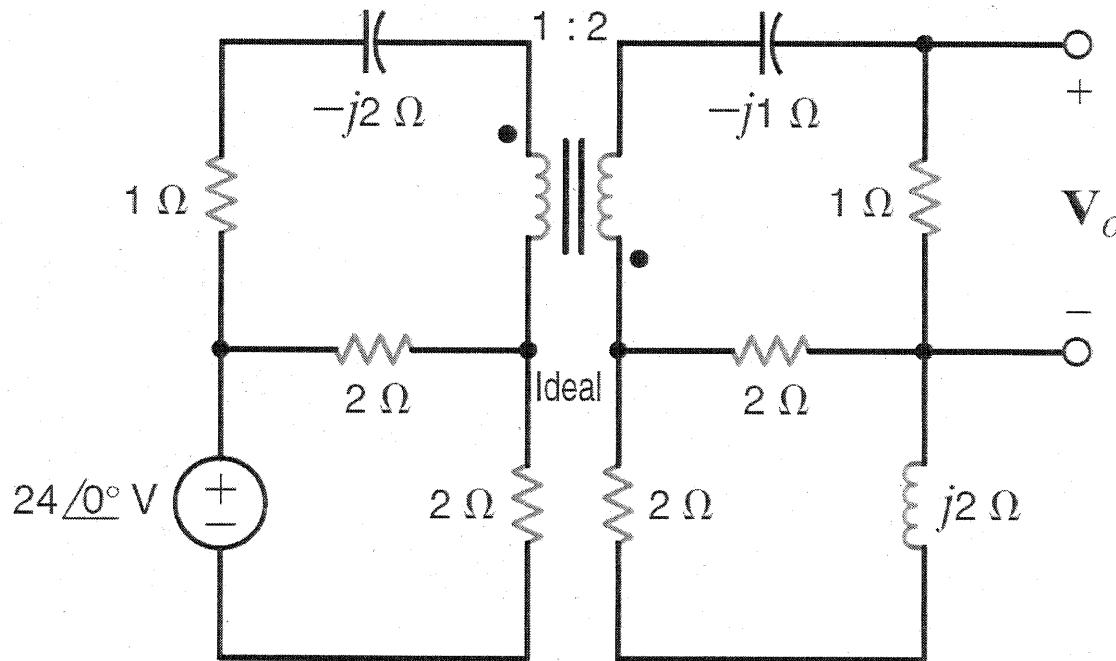
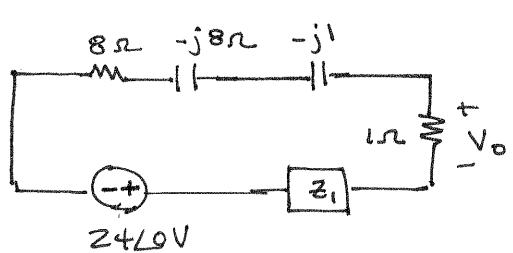
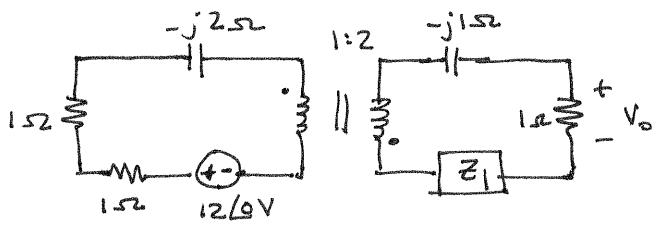


Figure P10.63

SOLUTION: First, we simplify,

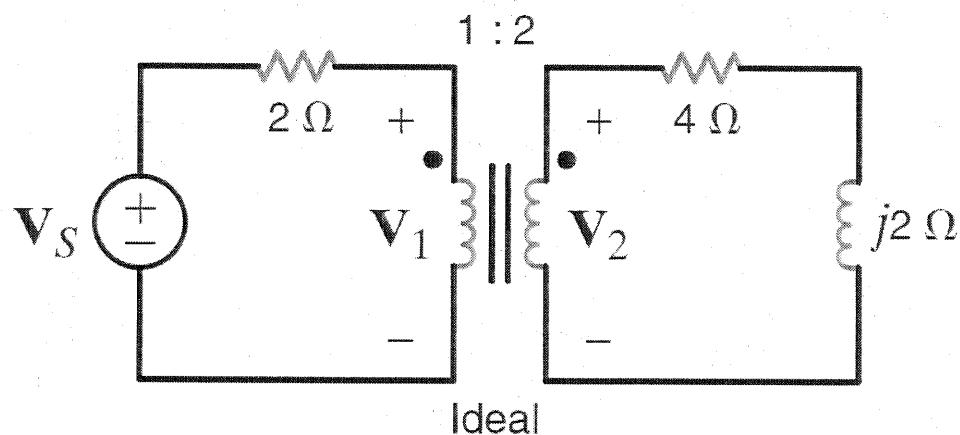
$$\begin{aligned}
 & \text{Initial circuit: } 24\angle 0^\circ \text{ V source, } 1\Omega \text{ resistor, } 2\Omega \text{ resistor, } -j2\Omega \text{ capacitor.} \\
 & \text{Simplify: } 24\angle 0^\circ \text{ V source, } \frac{2\Omega}{2\Omega} = 1\Omega \text{ resistor, } 12\angle 0^\circ \text{ A current, } 2\Omega \text{ resistor, } -j2\Omega \text{ capacitor.} \\
 & \text{Final simplified circuit: } 12\angle 0^\circ \text{ A current, } 1\Omega \text{ resistor, } 12\angle 0^\circ \text{ V source, } 2\Omega \text{ resistor, } -j2\Omega \text{ capacitor.} \\
 & \text{Calculate } Z_1: \\
 & Z_1 = \frac{Z(z+jz)}{4+j2} = 12 + j0.4 \Omega
 \end{aligned}$$



$$V_o = \frac{-24\angle 0^\circ (1)}{8+1 - j(8+1) + Z_1}$$

$$\boxed{V_o = 1.80 \angle -140^\circ \text{ V}}$$

- 10.64** Determine the input impedance seen by the source in the circuit in Fig. P10.64.



**Figure P10.64**

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SOLUTION:

$$Z_{in} = V_s / \pm = 3 + j0.5 \Omega$$

$Z_{in} = 3 + j0.5 \Omega$

- 10.65 Determine the input impedance seen by the source in the circuit in Fig. P10.65. **CS**

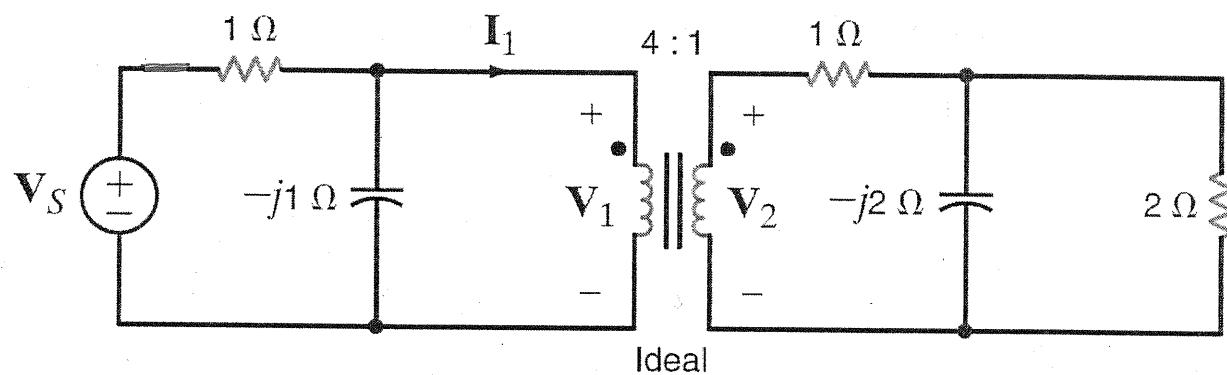
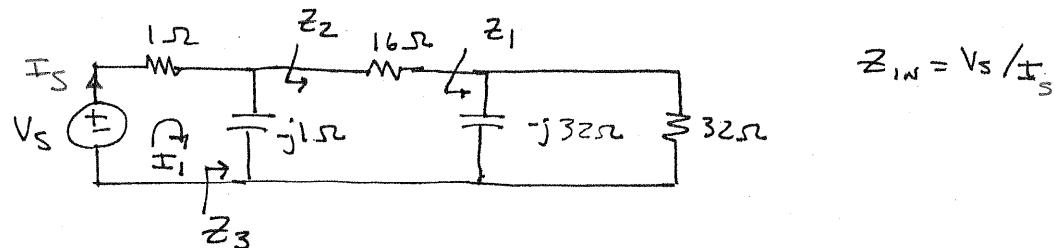


Figure P10.65

SOLUTION:



$$z_1 = \frac{32(-j32)}{32-j32} = 16-j16 \Omega \quad z_2 = 16 + z_1 = 32-j16 \Omega$$

$$z_3 = \frac{-j1(z_2)}{-j1+z_2} = 0.987 \angle -88.6^\circ \Omega \quad z_{in} = 1 + z_3$$

$$z_{in} = 1.42 \angle -43.9^\circ \Omega$$

**10.66** Determine the input impedance seen by the source in the network shown in Fig. P10.66.

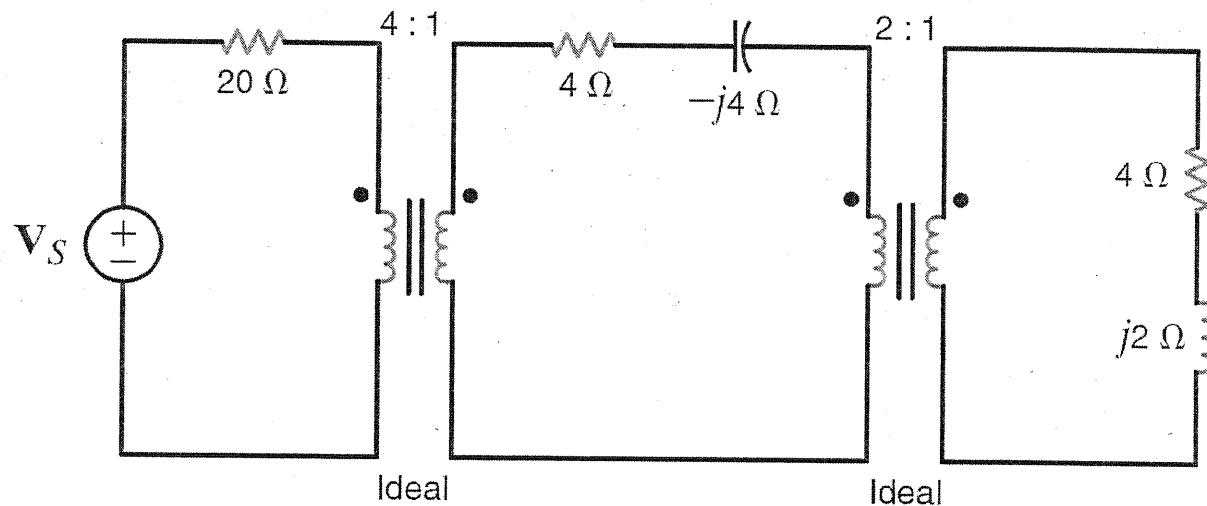
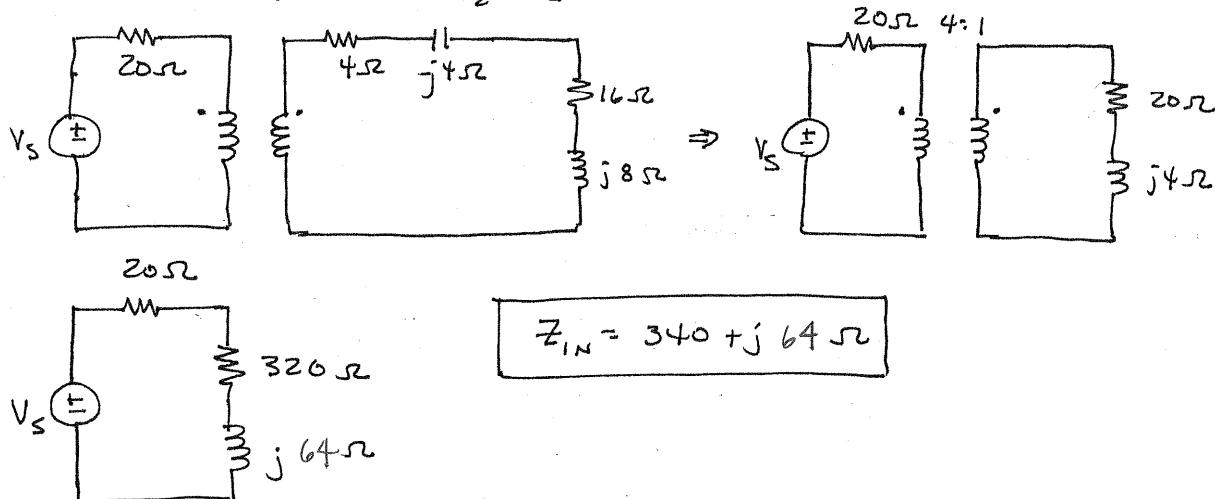


Figure P10.66

SOLUTION:  $n_1 = \frac{V_s}{4}$      $n_2 = \frac{1}{2}$



- 10.67 Determine the input impedance seen by the source in the network shown in Fig. P10.67.

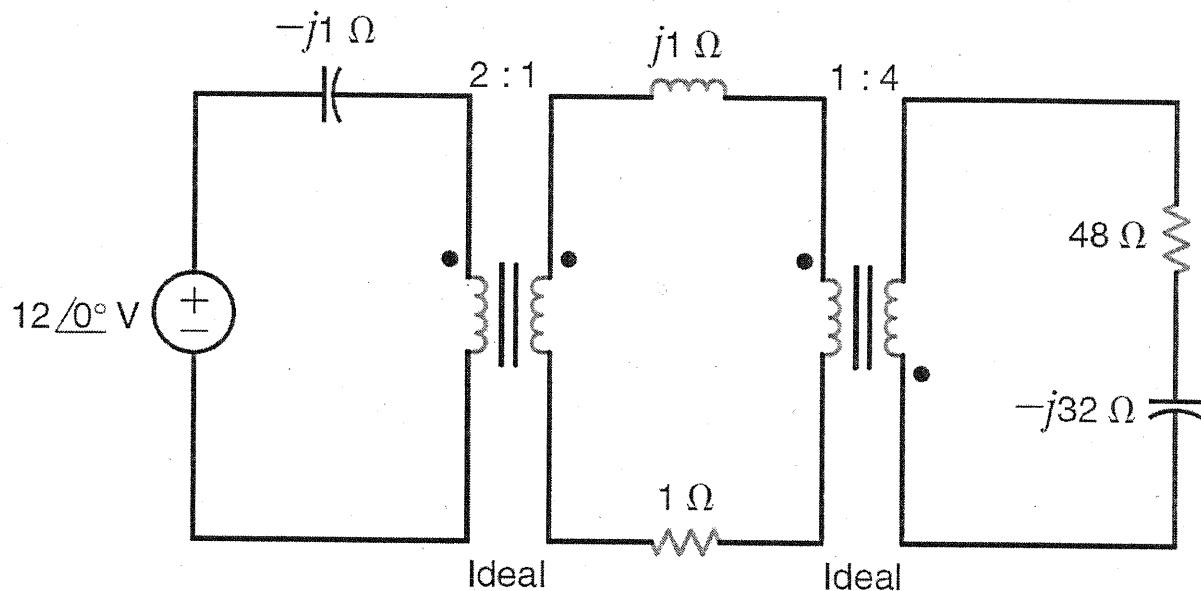
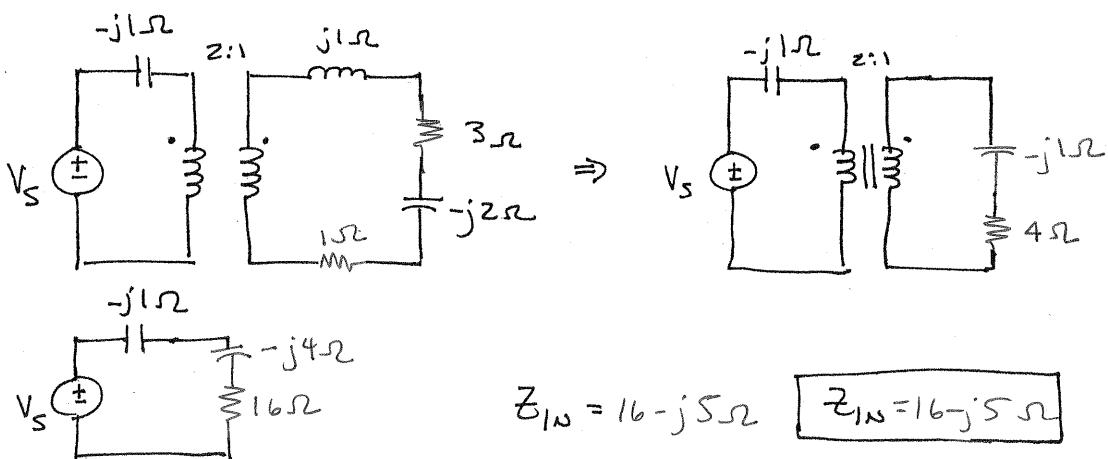
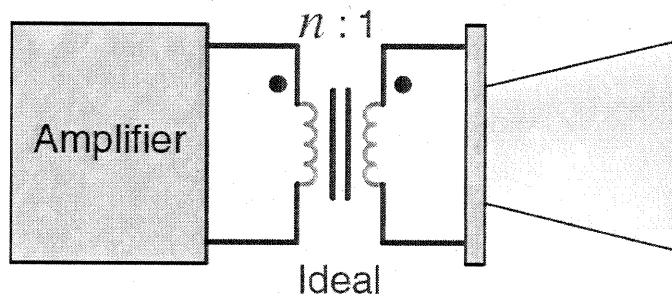


Figure P10.67

SOLUTION:



- 10.68** The output stage of an amplifier in an old radio is to be matched to the impedance of a speaker, as shown in Fig. P10.68. If the impedance of the speaker is  $8 \Omega$  and the amplifier requires a load impedance of  $3.2 \text{ k}\Omega$ , determine the turns ratio of the ideal transformer.



**Figure P10.68**

**SOLUTION:**

$$Z_{\text{SPEAKER}} = 8\Omega$$

$$Z_{\text{SPEAKER}} = n^2 Z_{\text{LOAD}}$$

$$n^2 = \frac{3200}{8} = 400$$

$$n = 20$$

10.69 Determine  $V_S$  in the circuit in Fig. P10.69.

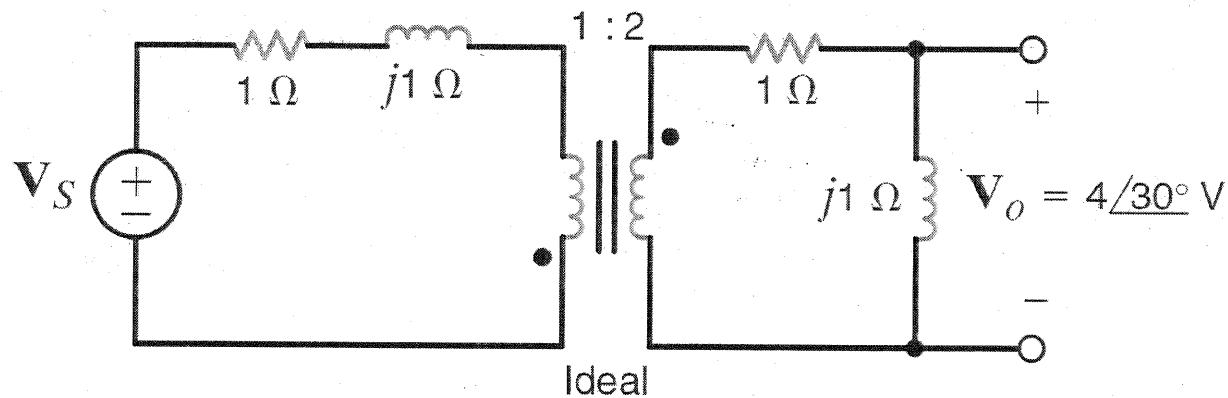
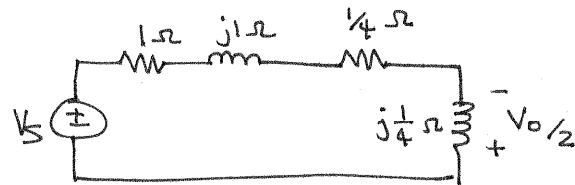


Figure P10.69

SOLUTION:



$$-\frac{V_o/2}{V_S} = \frac{j\frac{1}{4}}{1 + \frac{1}{4} + j(1 + \frac{1}{4})} = 0.1 + j0.1$$

$$V_S = -\frac{z \angle 30}{0.1 + j0.1}$$

$$V_S = 14.14 \angle 165^\circ V$$

10.70 Determine  $I_S$  in the circuit in Fig. P10.70.

**PSV**

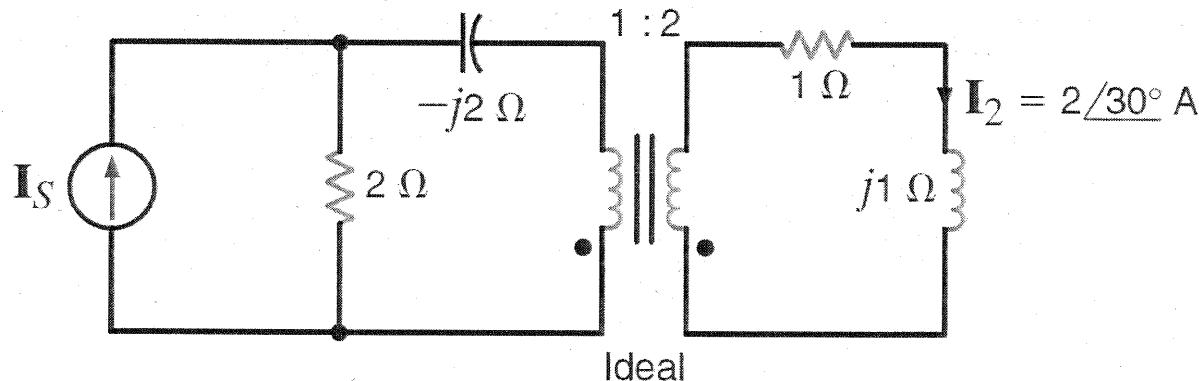
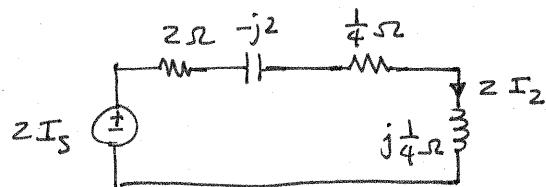


Figure P10.70

SOLUTION:



$$z I_2 = \frac{z I_s}{z .25 - j 1.75} = 4 \angle 30^\circ$$

$$I_s = 2 \angle 30^\circ (2.25 - j 1.75)$$

$$I_s = 5.70 \angle -7.87^\circ A$$

- 10.71 Given that  $\mathbf{V}_o = 48 \angle 30^\circ \text{ V}$  in the circuit shown in Fig. P10.71, determine  $\mathbf{V}_s$ . CS

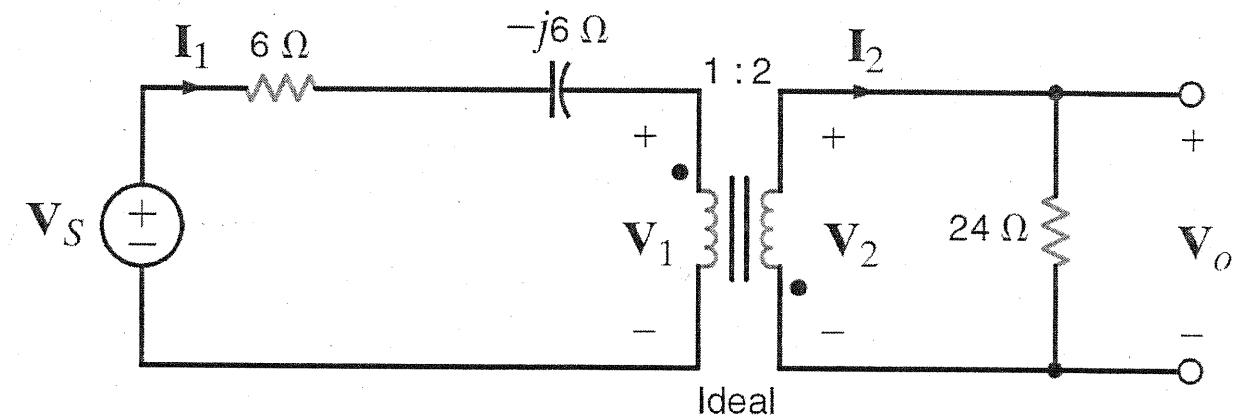


Figure P10.71

SOLUTION:

$$\mathbf{V}_2 = \mathbf{V}_o = 48 \angle 30^\circ \quad \mathbf{V}_1 = -\frac{\mathbf{V}_2}{2} = -24 \angle 30^\circ = 24 \angle -150^\circ \text{ V}$$

$$\mathbf{I}_2 = \mathbf{V}_o / 24 = 2 \angle 30^\circ \text{ A} \quad \mathbf{I}_1 = -2 \mathbf{I}_2 = -4 \angle 30^\circ \text{ A} = 4 \angle -150^\circ \text{ A}$$

$$\mathbf{V}_s = \mathbf{I}_1 (6 - j6) + \mathbf{V}_1$$

$$\mathbf{V}_s = 53.7 \angle -177^\circ \text{ V}$$

10.72 In the circuit in Fig. P10.72, if  $\mathbf{I}_x = 4 \angle 30^\circ \text{ A}$ , find  $\mathbf{V}_o$ .

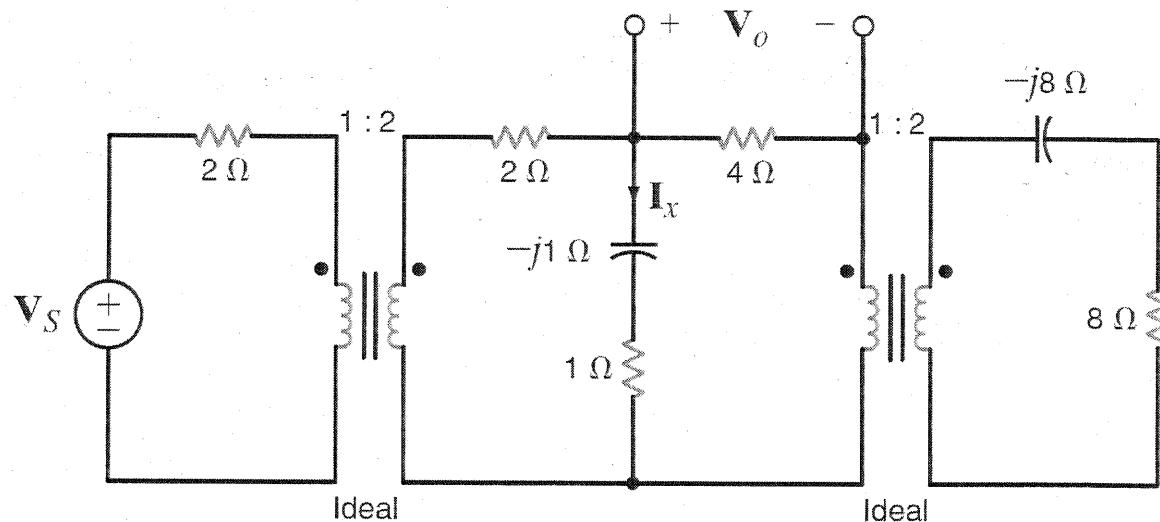
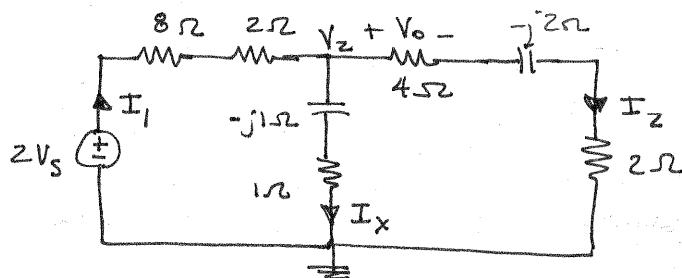


Figure P10.72

SOLUTION:



$$V_z = I_x (1 - j1) = 4\sqrt{2} \angle -15^\circ \text{ V}$$

$$I_z = \frac{V_z}{6 - j2} = 0.894 \angle 3.43^\circ \text{ A}$$

$$V_o = 4I_z = 3.58 \angle 3.43^\circ \text{ V}$$

10.73 In the network in Fig. P10.73, if  $\mathbf{I}_1 = 6/0^\circ \text{ A}$ , find  $\mathbf{V}_S$ .

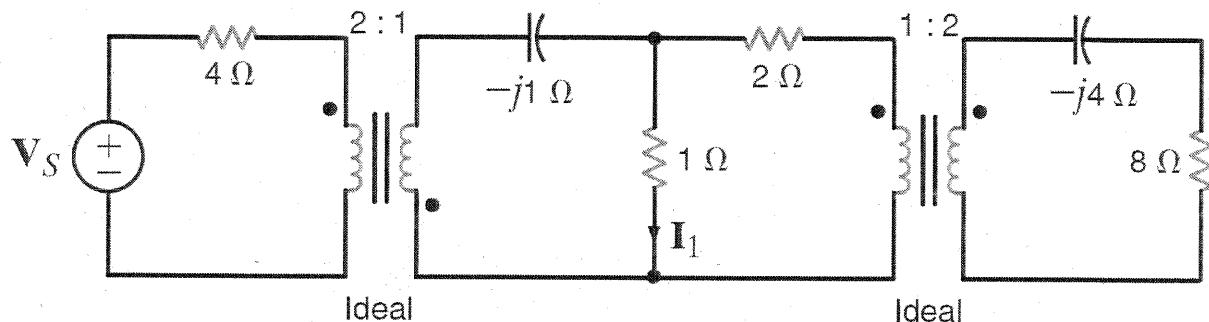
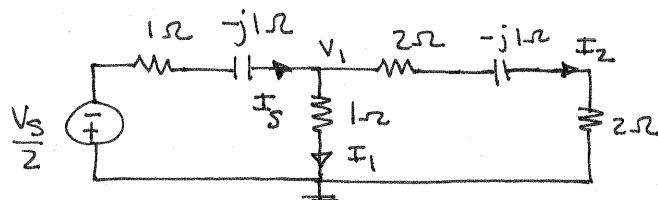


Figure P10.73

SOLUTION:



$$V_1 = 1 \cdot I_1 = 6/0^\circ \text{ V}$$

$$I_2 = V_1 / (4 - j1) = 1.46/14.0^\circ \text{ A}$$

$$I_S = I_2 + I_1 = 7.42/2.73^\circ \text{ A}$$

$$-\frac{V_S}{2} = I_S (1 - j1) + V_1 = 15.5/-27.1^\circ \text{ V}$$

$$\boxed{V_S = 30.9/153^\circ \text{ V}}$$

- 10.74** For maximum power transfer, we desire to match the impedance of the inverting amplifier stage in Fig. P10.74 to the  $50\text{-}\Omega$  equivalent resistance of the ac input source. However, standard op-amps perform best when the resistances around them are at least a few hundred ohms. The gain of the op-amp circuit should be  $-10$ . Design the complete circuit by selecting resistors no smaller than  $1\text{ k}\Omega$  and specifying the turns ratio of the ideal transformer to satisfy both the gain and impedance matching requirements.

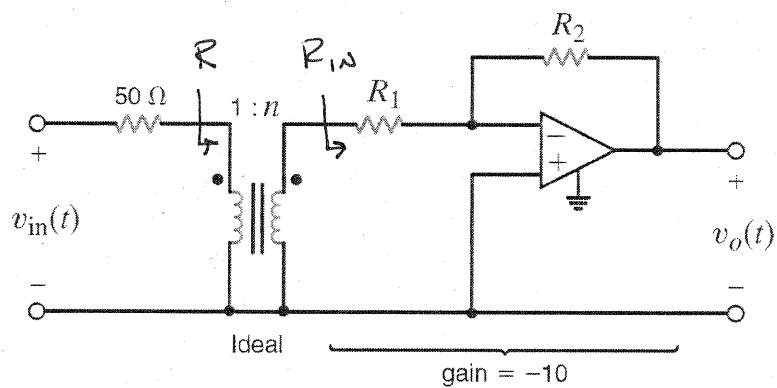


Figure P10.74

SOLUTION:

$$\text{gain} = -10 = -R_2/R_1 \quad \text{Arbitrarily select } R_1 = 5\text{ k}\Omega$$

yields  $R_2 = 50\text{ k}\Omega$

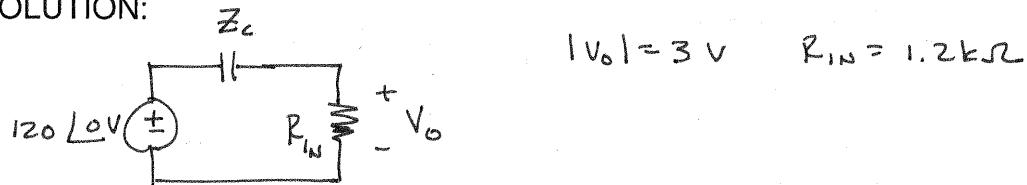
for op amp circuits  $R_{IN} = R_1 = 5\text{ k}\Omega$

$$R = R_{IN}/n^2 \Rightarrow n^2 = 5000/50 = 100 \Rightarrow n = 10$$

$R_1 = 5\text{ k}\Omega \quad R_2 = 50\text{ k}\Omega \quad n = 10$

**10.75** Digital clocks often divide a 60-Hz frequency signal to obtain a 1-second, 1-minute, or 1-hour signal. A convenient source of this 60-Hz signal is the power line. However, 120 volts is too high to be used by the low-power electronics. Instead, a 3-V, 60-Hz signal is needed. If a resistive voltage divider is used to drop the voltage from 120 V to 3 V, the heat generated will be unacceptable. In addition, it is costly to use a transformer in this application. Digital clocks are consumer items and must be very inexpensive to be a competitive product. The problem then is to design a circuit that will produce between 2.5 V and 3 V at 60 Hz from the 120-V ac power line without dissipating any heat or the use of a transformer. The design will interface with a circuit that has an input resistance of 1200 ohms.

SOLUTION:



$$|V_0| = 3 \text{ V} \quad R_{IN} = 1.2 \text{ k}\Omega$$

$$V_0 = \frac{120 \text{ V} R_{IN}}{R_{IN} + Z_c} \Rightarrow |V_0| = 3 = \frac{120 (1200)}{\sqrt{1200^2 + Z_c^2}}$$

$$Z_c^2 = 2.30 \times 10^9 \Rightarrow Z_c = -j47.98 \text{ k}\Omega$$

$$Z_c = -j\omega C = -j47.98 \text{ k}\Omega \quad C = 55.3 \text{ nF}$$

**10FE-1** In the network in Fig. 10PFE-1, find the impedance seen by the source. **cs**

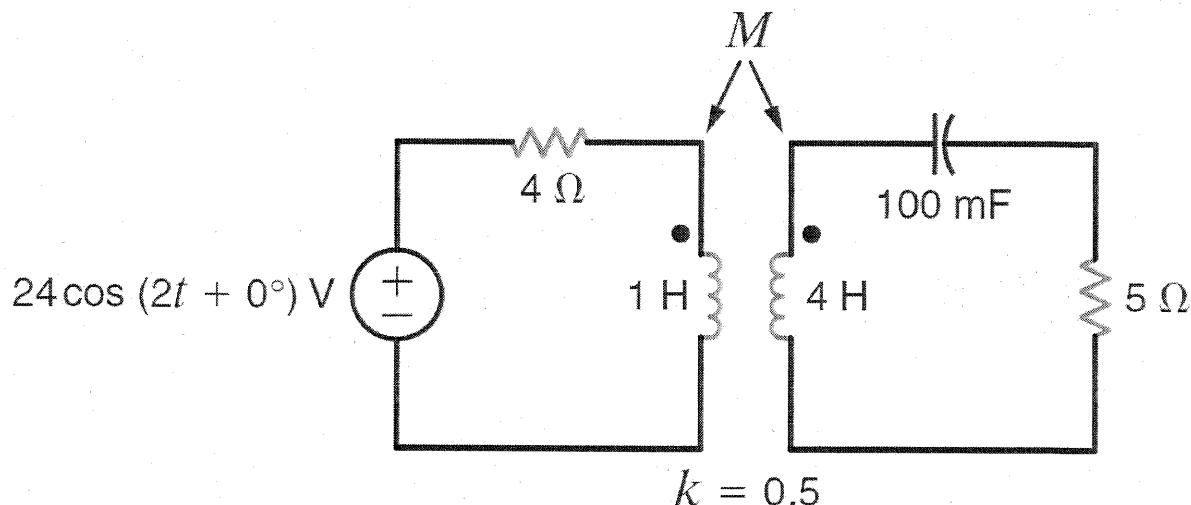
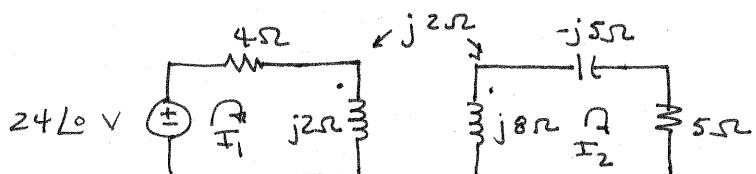


Figure 10PFE-1

SOLUTION:  $\omega = Z \text{ r/s}$        $M = k \sqrt{L_1 L_2} = 1 \text{ H}$



$$\begin{aligned} 24\angle 0^\circ &= I_1 (4+j2) - j2 I_2 \\ 0 &= -j2 I_1 + I_2 (5+j3) \end{aligned} \Rightarrow \begin{bmatrix} 4+j2 & -j2 \\ -j2 & 5+j3 \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 24 \\ 0 \end{bmatrix}$$

$$I_1 = 4.92 \angle -19.7^\circ \text{ A}$$

$$Z_{IN} = \frac{24\angle 0^\circ}{I_1}$$

$$Z_{IN} = 4.87 \angle 19.7^\circ \Omega$$

**10FE-2** In the circuit in Fig. 10PFE-2, select the value of the transformer's turns ratio  $n = N_2/N_1$  to achieve impedance matching for maximum power transfer. Using this value of  $n$ , calculate the power absorbed by the  $3\Omega$  resistor.

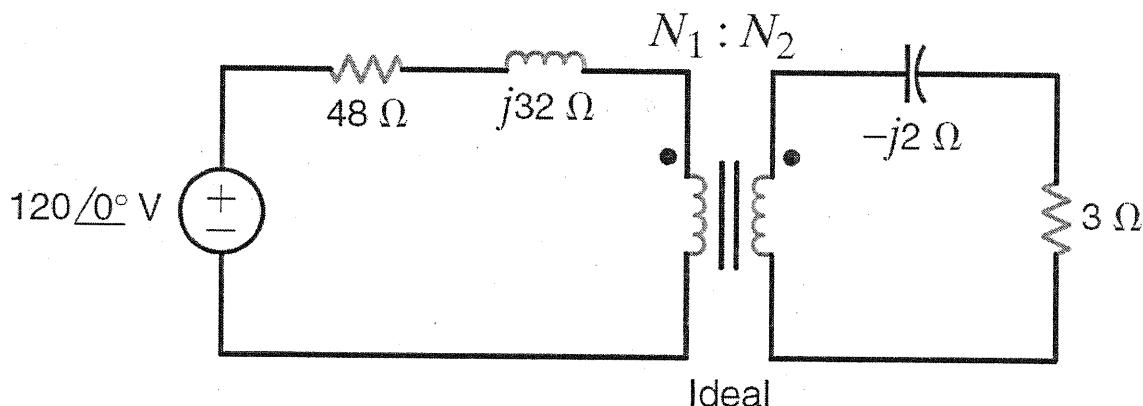
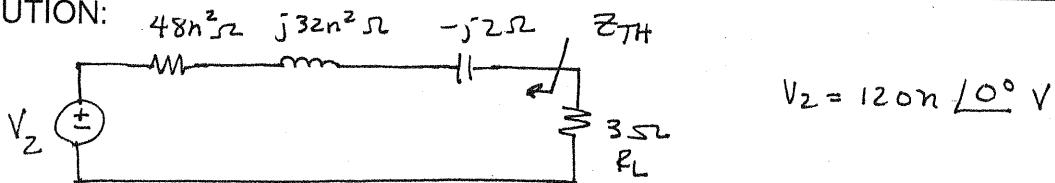


Figure 10PFE-2

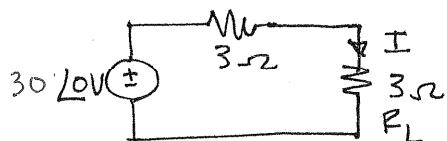
SOLUTION:



For max. power transfer,  $Z_{TH} = 3\Omega$

$$Z_{TH} = 48n^2 + j(32n^2 - 2) = 3$$

$$\text{if } n = 1/4, \quad Z_{TH} = 48/16 + j(32/16 - 2) = 3\Omega \quad \checkmark$$



$$I = \frac{30∠0}{6} = 5∠0 A$$

$$P_L = \frac{1}{2} I_m^2 R_L$$

$P_L = 37.5 W$
$n = 1/4$

- 10FE-3** In the circuit in Fig. 10FE-3, select the turns ratio of the ideal transformer that will match the output of the transistor amplifier to the speaker represented by the  $16\text{-}\Omega$  load. **CS**

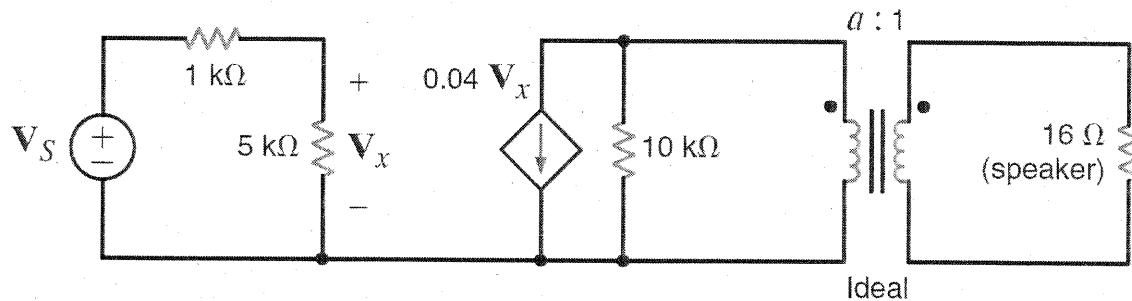


Figure 10PFE-3

**SOLUTION:**

Find Root of 2nd Order

$$\text{Circuit diagram: } V_s \text{ in series with } 1\text{k}\Omega \text{ to } V_x, \text{ then } 5\text{k}\Omega \text{ to ground.}$$

$$\text{Dependent source: } \frac{4V_x}{100} \text{ in series with } 10\text{k}\Omega \text{ to } V_{oc}.$$

$$V_{oc} = \left(\frac{5}{6} V_s\right) \left(-\frac{4}{100}\right) (10^4)$$

$$\text{Circuit diagram: } V_s \text{ in series with } 1\text{k}\Omega \text{ to } V_x, \text{ then } 5\text{k}\Omega \text{ to ground.}$$

$$\text{Dependent source: } \frac{4V_x}{100} \text{ in parallel with } 10\text{k}\Omega \text{ to } I_{sc}.$$

$$I_{sc} = \left(\frac{5}{6} V_s\right) \left(-\frac{4}{100}\right)$$

$$R_{out} = V_{oc} / I_{sc} = 10\text{k}\Omega$$

$$16a^2 = 10^4 \Rightarrow a = 25$$